



Real-Time Feedback on Beam Parameters

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- Stability Requirements
- Medium and Long Term Stability
- Beam-Based Feedback Design:
 - Space & Time Domain
 - Optimal Controller Design and Non-Linear Control
 - Cross-Dependability and -Constraints between FB loops

Disclaimer: Details on instrumentation covered in other presentation

- Accelerators can be grouped into three groups
 - **Light Sources:** (list not exhaustive¹⁻³)
ALBA, ANKA, ALS, APS, BSRF, BESSY, CLS, DELTA, ELETTRA, ESRF, INDUS1, LNSLS, SLS, DIAMOND, SOLEIL, SPEAR3, Spring-8, Super-ACO...
 - mostly orbit and energy (radial steering) only
 - **Lepton Collider:** LEP⁴, PEP-II⁵, KEK-B
 - orbit and tune feedback (mostly during ramp)
 - **Hadron Collider:** Hera, LHC, RHIC, Tevatron
 - mostly slow orbit feedback, except:
 - Hera: Orbit, Tune
 - RHIC: Tune⁶/Coupling, Chromaticity⁷
 - LHC: Orbit/Energy, Tune/Coupling, Chromaticity, ...

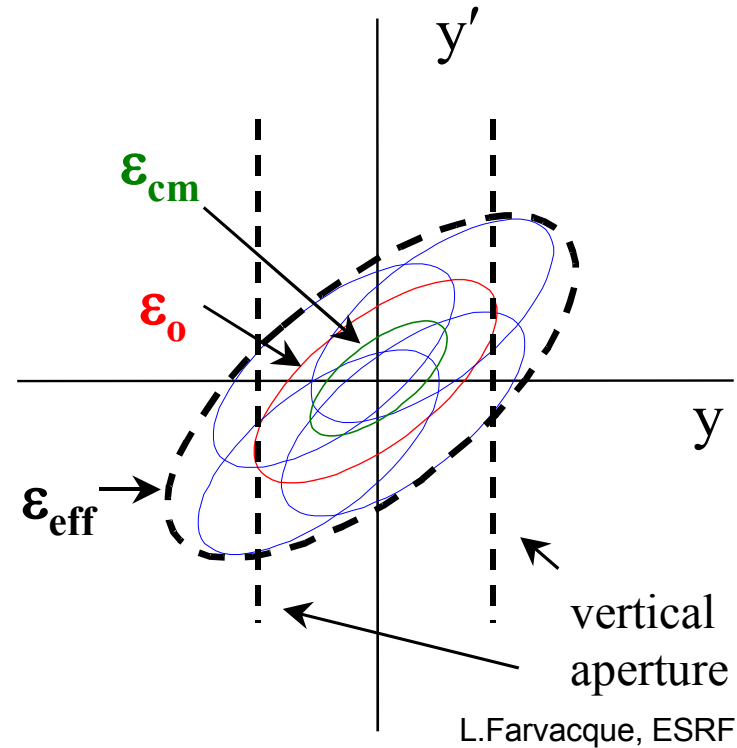
- Main requirements for orbit stability⁸:

- Effective emittance preservation
(τ_d sampling/integration time, τ_f fluctuation time)

$$\tau_d \gg \tau_f: \quad \epsilon_{eff} = \epsilon_0 + \epsilon_{cm}$$

$$\tau_d \ll \tau_f: \quad \epsilon_{eff} \approx \epsilon_0 + 2\sqrt{\epsilon_0 \epsilon_{cm}} + \epsilon_{cm}$$

- Minimisation of coupling
(vertical orbit in sextupoles)
- Minimisation of spurious dispersion
(vertical orbit in quadrupoles)
- Collider Luminosity and collision point stability (in case of two separated rings)



L.Farvacque, ESRF

$$L = L_0 \cdot \exp \left\{ \frac{(\bar{x} - x)^2}{2\sigma_x^2} + \frac{(\bar{y} - y)^2}{2\sigma_y^2} \right\} \cdot 1 / \sqrt{1 + \left(\frac{\theta_c \sigma_z}{2\sigma_{x/y}} \right)^2} \cdot \dots$$

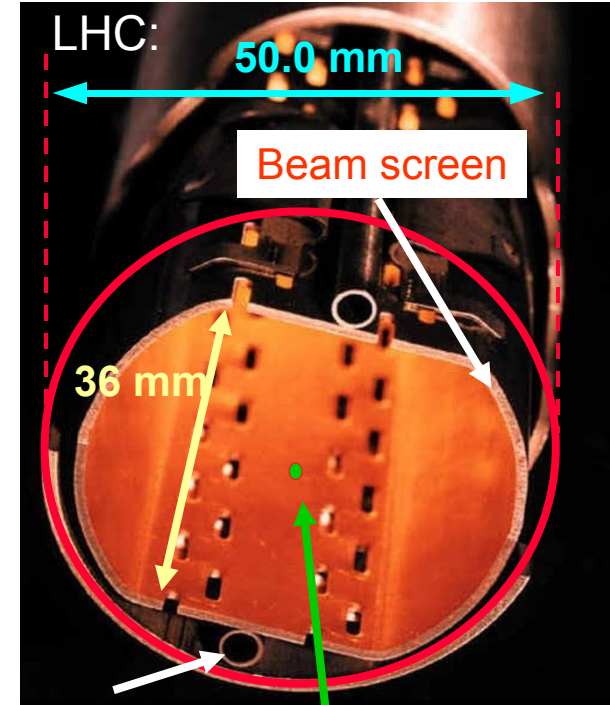
→ Nearly all 3rd generation light-sources deploy at least orbit/energy feedbacks¹⁻³

- Traditional requirements on beam stability...

... to keep the beam in the pipe!

- Increased stored intensity and energy:
 - sufficient to quench all magnets and/or to cause serious damage⁹

- Requirements depend on:
 - Capability to control particle losses in the machine
 - Machine protection & Collimation
 - Quench prevention
 - Commissioning and operational efficiency



Cooling channel (He)

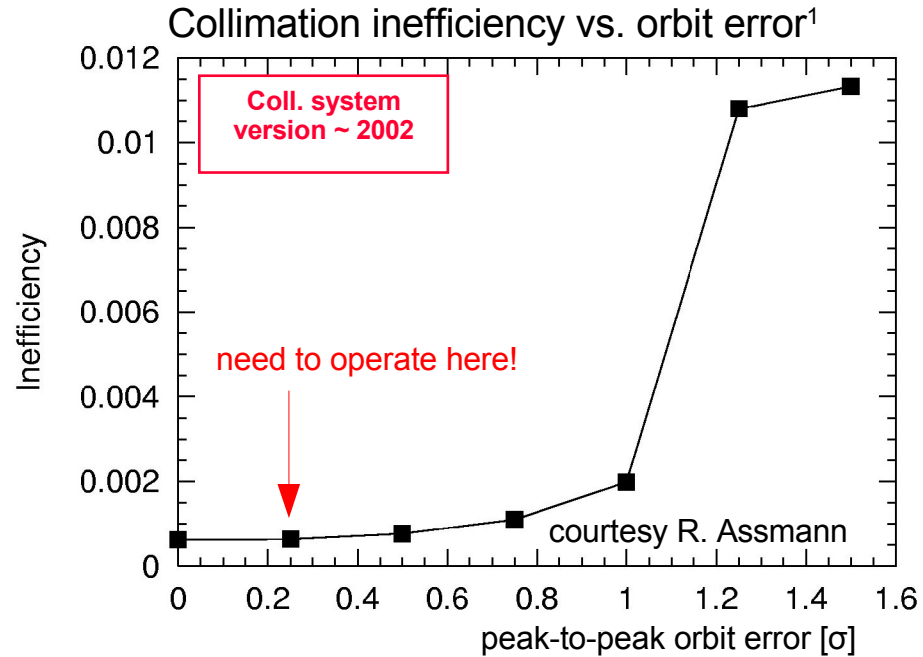
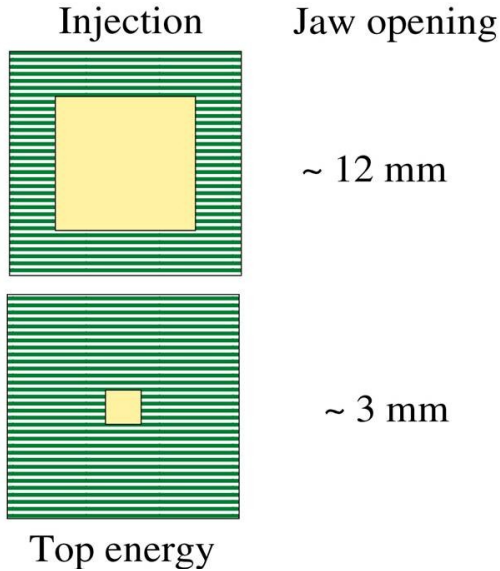
Beam 3σ envel.
~ 1.8 mm @ 7 TeV

Hadron Collider Requirements

LHC Collimation System and Closed Orbit



10 mm



- LHC Collimation System, $N_{\max} \approx 5 \cdot 10^{14}$ protons/beam (nominal)

- required collimation inefficiency^{10,11}:

$$\eta = \frac{\text{number of particles escaping collimation}}{\text{number of particles impacting collimation}}$$

→ LHC: $\eta < 0.001$

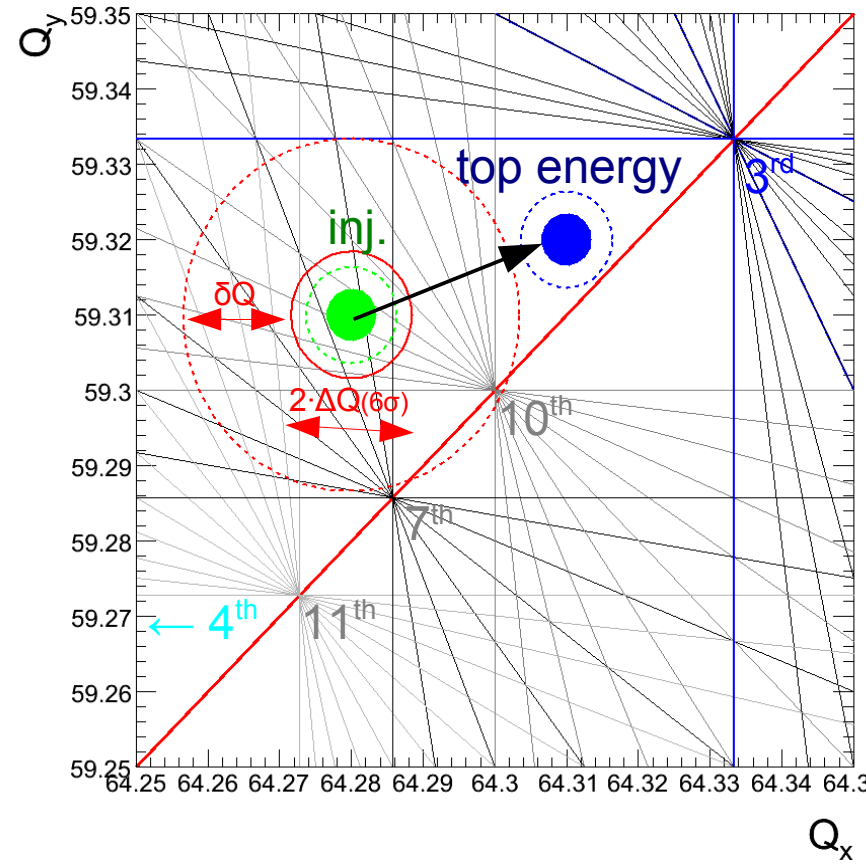
- Orbit stability requirement better than $\sigma/6 < \sim 25 \mu\text{m}$ at collimator jaws
- Several other similar and distributed requirements:
 - local \approx global requirements¹³

Hadron Collider Requirements on Tune and Chromaticity

- Lepton machines: $\delta Q \sim 10^{-2} \dots 10^{-3}$
 - avoid up to $\sim 3^{\text{rd}}$ order resonance

- Hadron machines:
 - negligible synch. radiation damping
 - large tune footprints
 - avoid up to 12^{th} order resonances

- Example LHC:
 - Tune spread (LHC) $\Delta Q|_{\text{av}} \approx 1.15 \cdot 10^{-2}$
(fixed by available space in Q-diagram)
 - $\delta Q \leq 0.003 \dots 0.001$ (nominal)
 - Chromaticity (SPS: $\Delta p/p \approx 2.8 \cdot 10^{-4}$)



- allowed max lin. chromaticity^{14,15} (5-6 σ , first order):

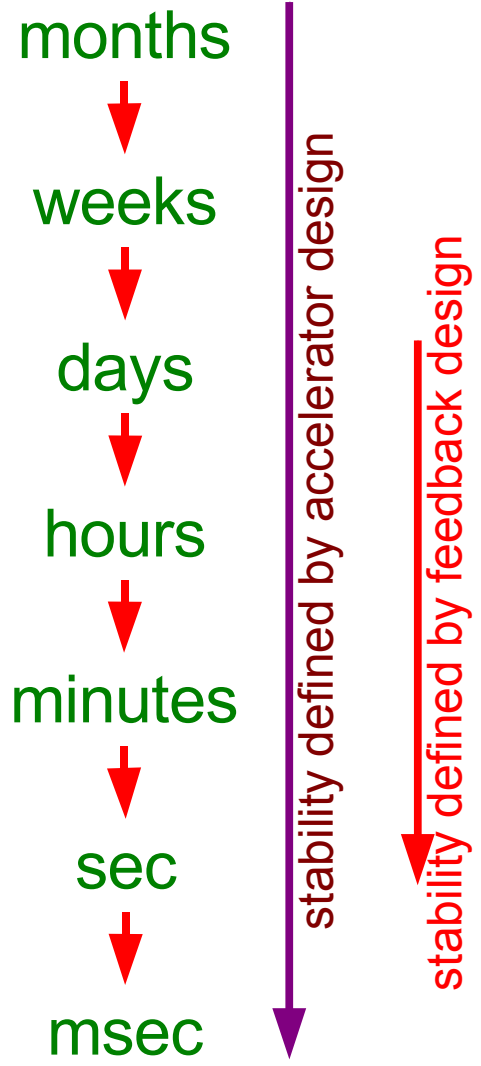
$$Q'_{max} \propto \frac{\Delta Q_{av}}{\Delta p/p}$$

$$\rightarrow Q'_{max} \approx 2 \pm 1 \ \& \ Q' > 0$$

(expected drifts¹: $\Delta Q' \approx 140$)

■ ...can be grouped into:

- **Environmental sources:**
(mostly propagated through quadrupoles/girders)
 - temperature and pressure changes,
 - ground motion, tides,
 - cultural noise
- **Machine inherent sources:**
 - decay and snap-back of multipoles,
 - cooling liquid flow, pumps/ventilation vibrations
 - eddy currents
 - changes of machine optics (final focus)
- **Machine element failures:**
 - corrector circuits (LHC: 1300++ circuits)



- Feedbacks perform well on short to medium term time scales
- Corrector circuit noise and systematics
 - Hysteresis, eddy-currents, ADC quantisation noise, element failure
 - trend: 10/12 ADC bits \rightarrow 16 ADC bits²⁶ \rightarrow 18/20 ADC bits^{27,28,29}
- Beam instrumentation noise and its systematic
 - Dependence on bunch length and intensity (charge)
 - Thermal expansion of girders, drift of electronics
 - mechanically decoupling/stiffening of BPM girders (Invar, Carbon Fibre)
 - extensive temperature stabilisation:
 - tunnel and cooling water
 - discrete photon absorbers
 - water cooled vacuum chambers
 - 'top-up' operation^{1,2}

thermal exp. of steel:
1m of steel, 1°C:
 $\Delta x \approx 10.. 17 \mu\text{m}$

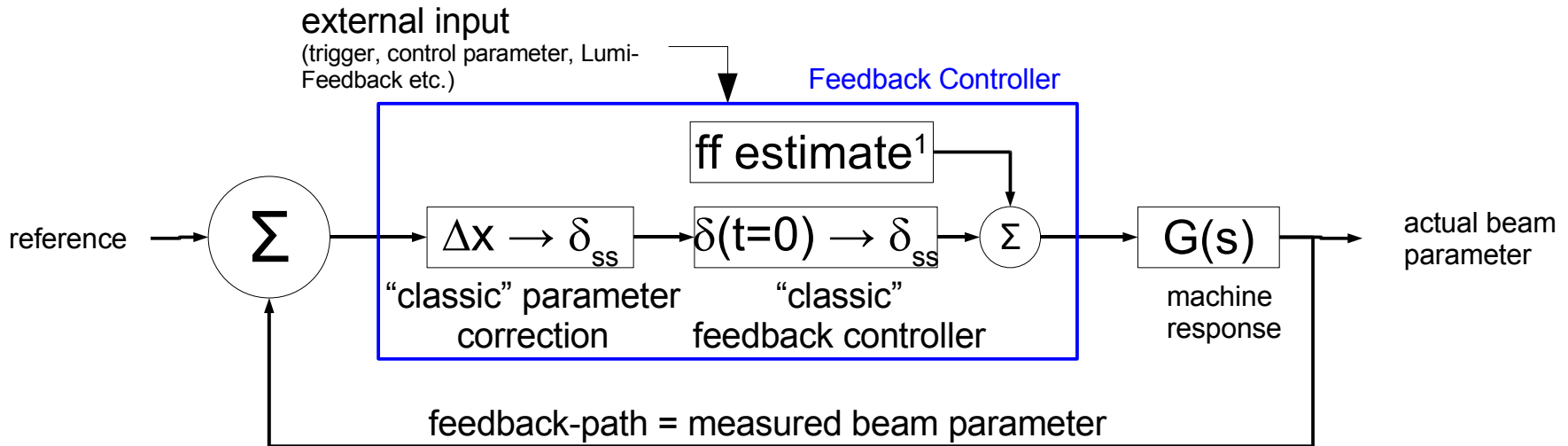
e.g. SLS²⁵:
electronics: $\pm 0.1 \text{ }^\circ\text{C}$
experimental hall: $\pm 1.0 \text{ }^\circ\text{C}$

- Feedback controller usually decomposed into three stages:

- 1 Compute steady-state corrector settings $\vec{\delta}_{ss} = (\delta_1, \dots, \delta_n)$ based on measured parameter shift $\Delta x = (x_1, \dots, x_n)$ that will move the beam to its reference position for $t \rightarrow \infty$.
- 2 Compute a $\vec{\delta}(t)$ that will enhance the transition $\vec{\delta}(t=0) \rightarrow \vec{\delta}_{ss}$
- 3 Feed-forward: anticipate and add deflections $\vec{\delta}_{ff}$ to compensate changes of well known and properly described¹ sources:

space domain

time domain



¹ properly described = accurate & fast real-time model of the source

Space-Domain: No “black feedback magic”

- Effects on orbit, Energy, Tune, Q' and C- can essentially be cast into matrices:

$$\Delta \vec{x}(t) = \underline{R} \cdot \vec{\delta}(t) \quad \text{with} \quad R_{ij} = \frac{\sqrt{\beta_i \beta_j}}{2 \sin(\pi Q)} \cdot \cos(\Delta \mu_{ij} - \pi Q) + \frac{D_i D_j}{C(\alpha_c - 1/\gamma^2)}$$

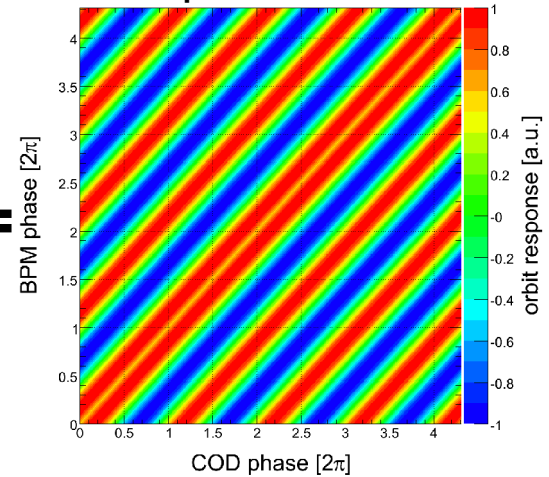
matrix multiplication

- similar for other parameters:

- Tune, Coupling,
- Chromaticity,
- Energy, ...

$$\begin{pmatrix} \Delta x_1 \\ \Delta x_2 \\ \dots \\ \Delta x_m \end{pmatrix}$$

orbit response matrix



$$\times \begin{pmatrix} \sigma_1 \\ \sigma_2 \\ \vdots \\ \sigma_n \end{pmatrix}$$

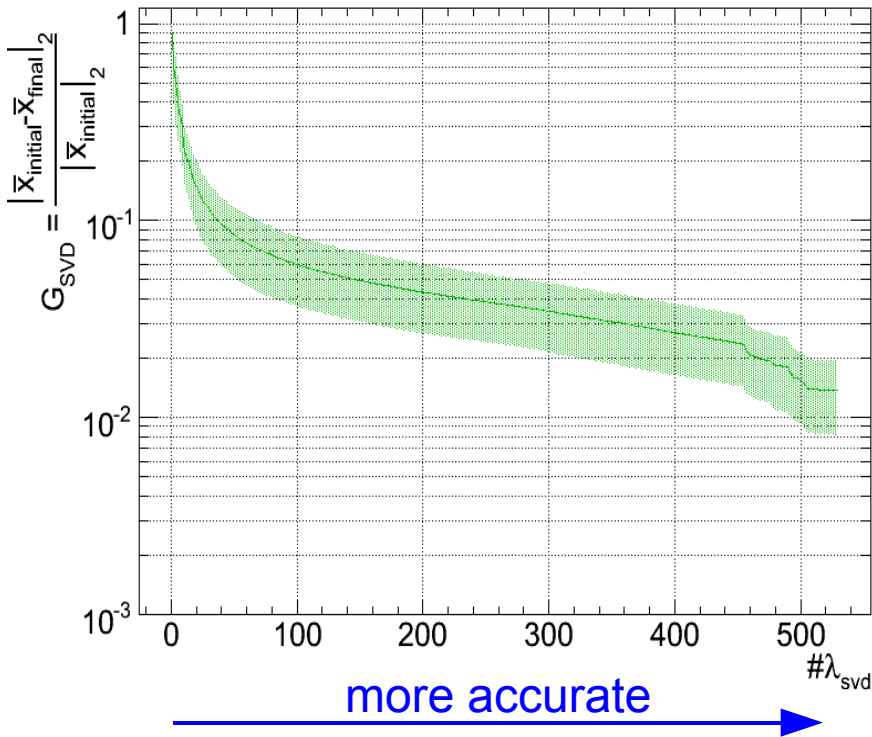
- Control consists essentially in inverting above matrices
 - Potential complications: **singularities** = over/under-constraint matrices, **noise**, **element failures**, spurious BPM offsets, calibrations errors, ...
 - Common Workhorse: Singular-Value-Decomposition (SVD¹⁶)
 - all light sources (& LHC) go in this direction!!

Intrinsic Trade-off of Parameter Controls

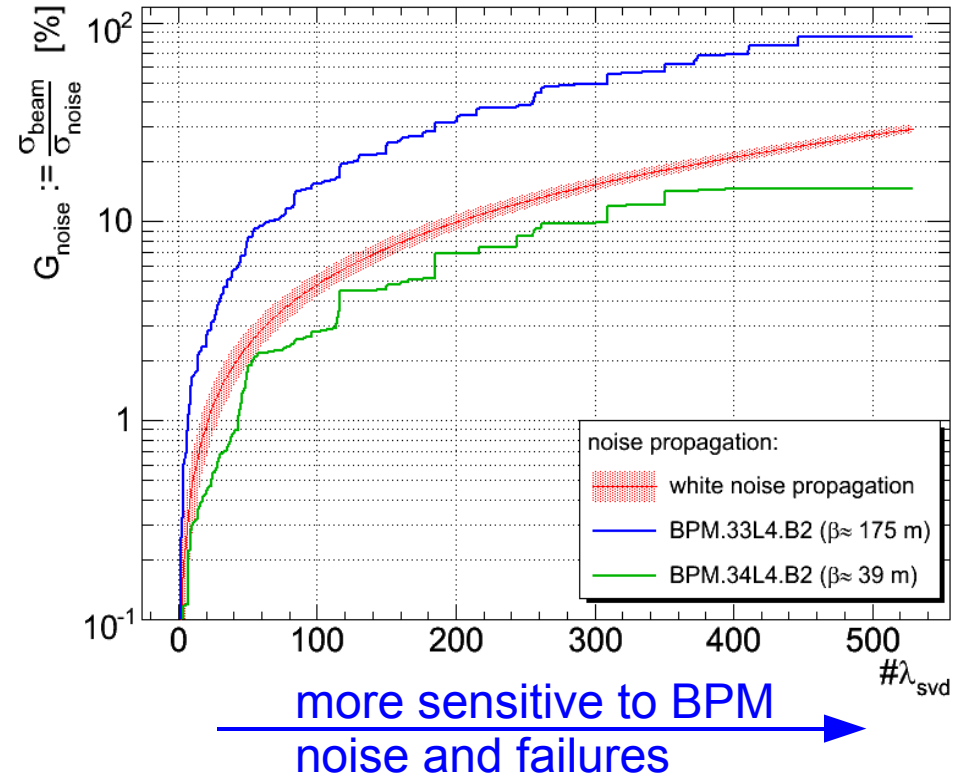
Example: SVD based LHC orbit correction

- Number of for the inversion used eigenvalues $\#\lambda_{\text{svd}}$ steers accuracy versus robustness of correction algorithm

- Orbit attenuation



- Sensitivity to BPM noise

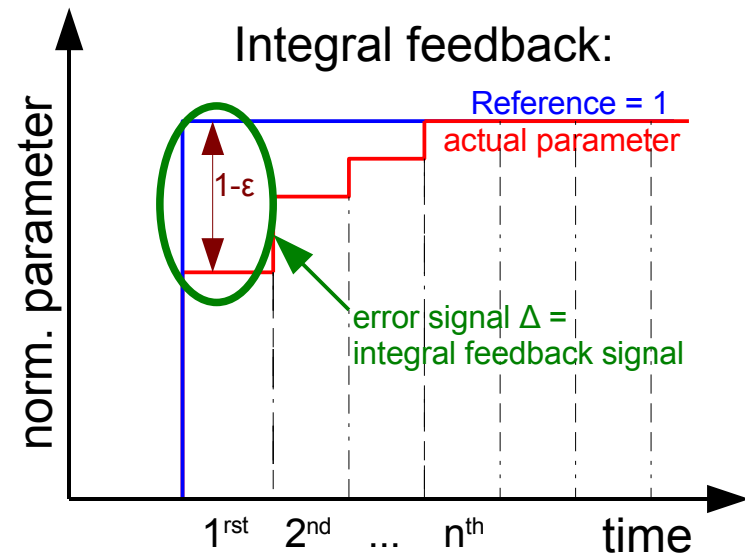
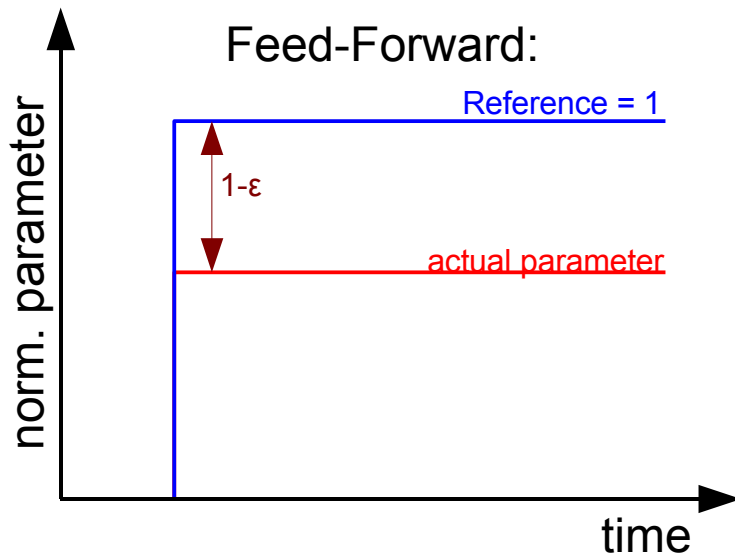


Lattice and Calibration Imperfections

Usage of Designed vs. Measured Beam response

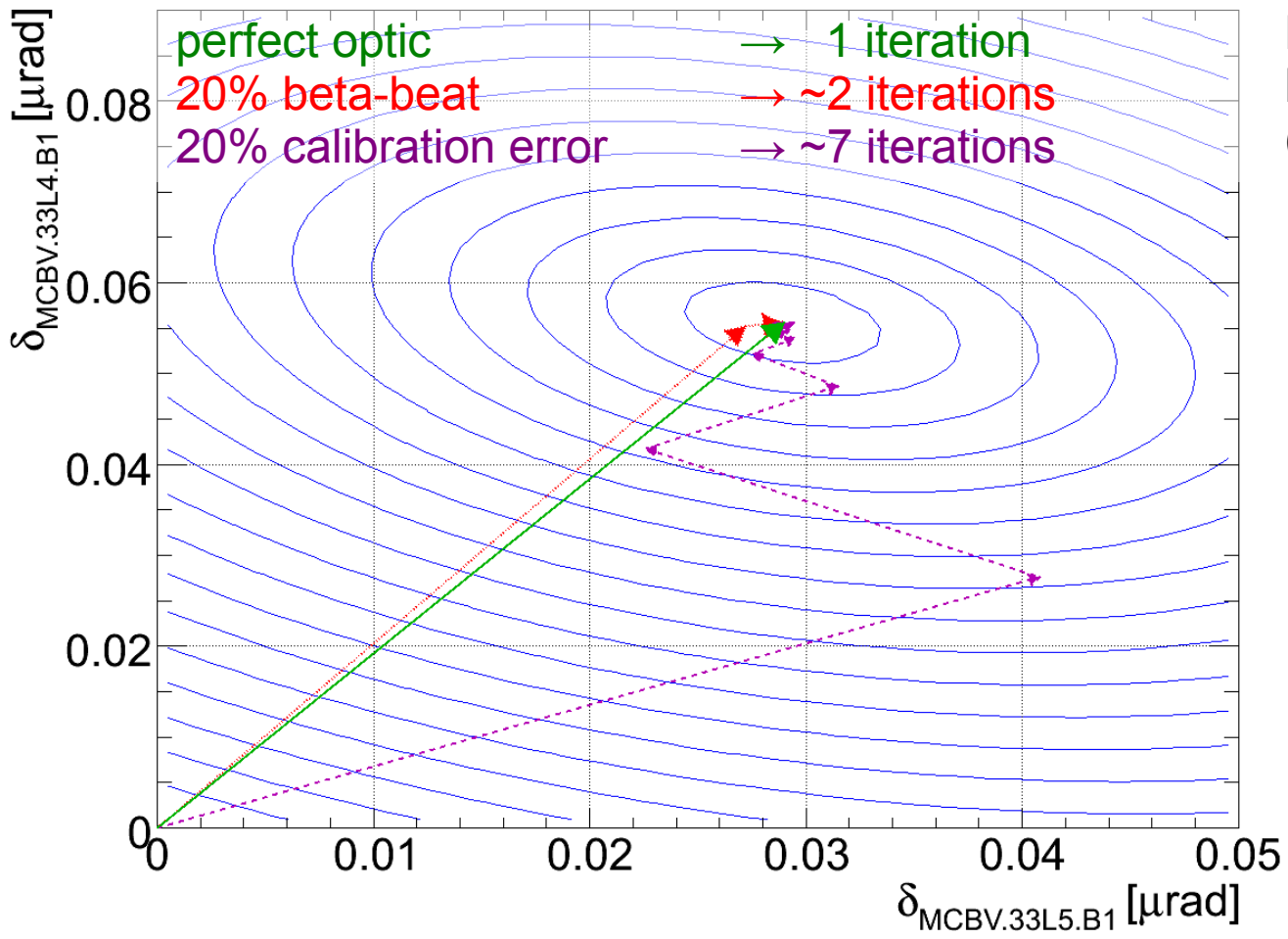
- Machine imperfections (beta-beat, hysteresis....), calibration errors and offsets can be translated into a steady-state ϵ_{ss} and scale error ϵ_{scale} :

$$\Delta x(s) = R_i(s) \cdot \delta_i \rightarrow \Delta x(s) = R_i(s) \cdot (\epsilon_{ss} + (1 + \epsilon_{scale}) \cdot \delta_i)$$



- Uncertainties and scale error of beam response function affects rather the convergence speed (= feedback bandwidth) than achievable stability

- SVD algorithm part of the class of gradient based minima searches



Example:
LHC arc
(FODO Lattice)

- Implication on optimal feedback sampling frequency:

- digital approx. of analogue system: $f_s > 10...20 \cdot f_{\text{bw}}$ $\xrightarrow{\text{imperf}}$ $f_s > 20...40++ \cdot f_{\text{bw}}$
- Example f_s/f_{bw} ratios: SLS: 40, ALS: 45, LHC: 25/50 (budget limited)

Time Domain: Optimal Controller Design

Youla's affine parameterisation

- PID/Controller design often regarded as specialists topic only - wrong!
- Youla^{17,18} showed that all stable closed loop controllers $D(s)$ can be written as:

$$D(s) = \frac{Q(s)}{1 - Q(s)G(s)} \quad (1)$$

- Example: first order system

$$G(s) = \frac{K_0}{\tau s + 1} \quad \text{with } \tau \text{ being the circuit time constant} \quad (2)$$

- Using for example the following ansatz:

$$Q(s) = F_Q(s) G^i(s) = \frac{1}{\alpha s + 1} \cdot \frac{\tau s + 1}{K_0} \quad (3)$$

- $F_Q(s)$ models the desired closed-loop response $\rightarrow T_0(s) = \frac{1}{\alpha s + 1}$
- $G^i(s)$ being the pseudo-inverse of the nominal plant $G(s)$

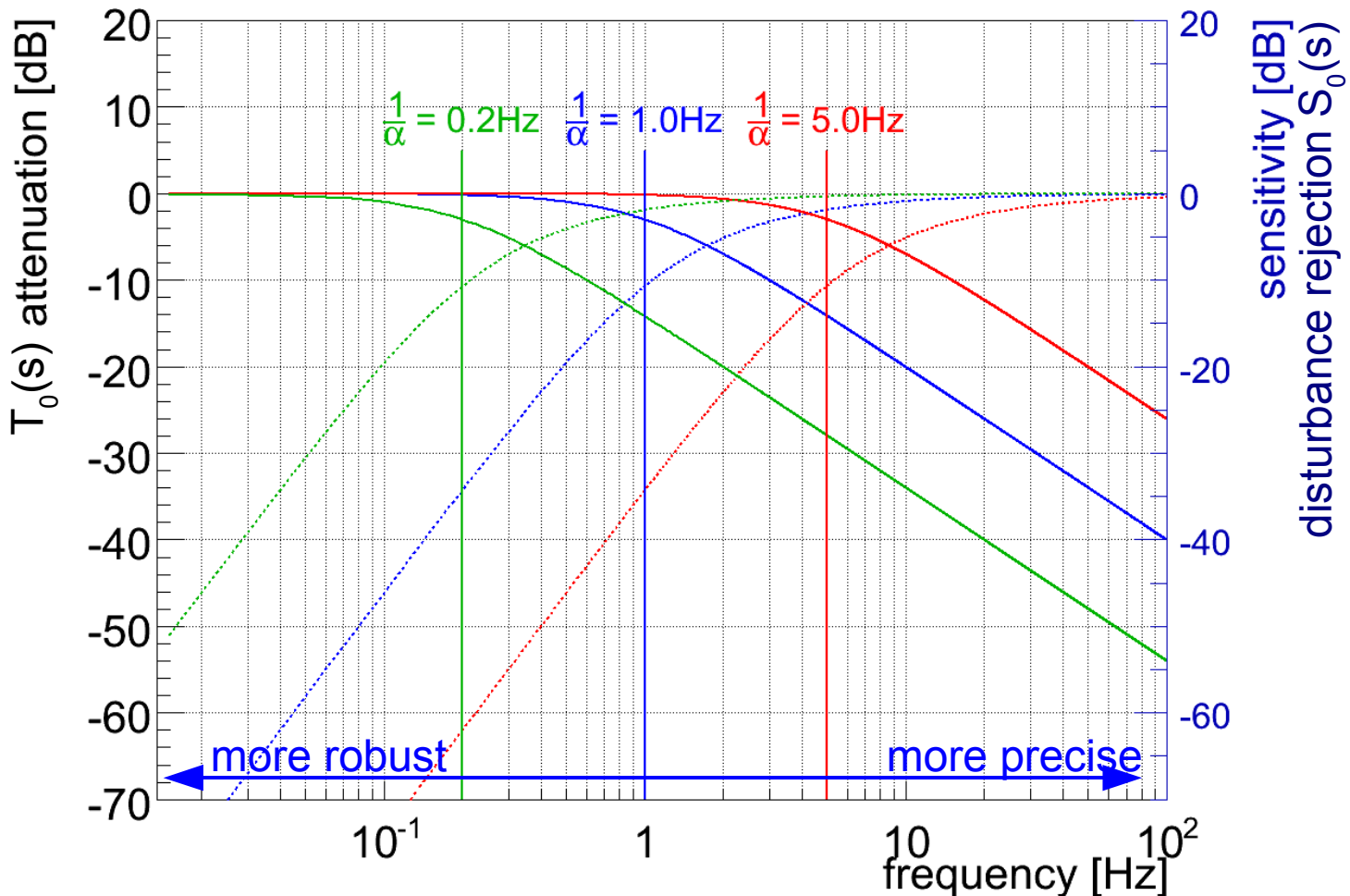
- (1)+(2)+(3) yields PI controller:

$$D(s) = K_p + K_i \frac{1}{s} \quad \text{with} \quad K_p = K_0 \frac{\tau}{\alpha} \wedge K_i = K_0 \frac{1}{\alpha}$$

Time Domain: Optimal Controller Design

Robust vs. Fast Closed Loop Response

- $\alpha > T \dots \infty$ facilitates the trade-off between speed and robustness
 - quantitative verification of loop stability w.r.t. noise, model errors ...
 - operator has to deal with one parameter \rightarrow enables simple adaptive gain-scheduling based on the operational scenario!

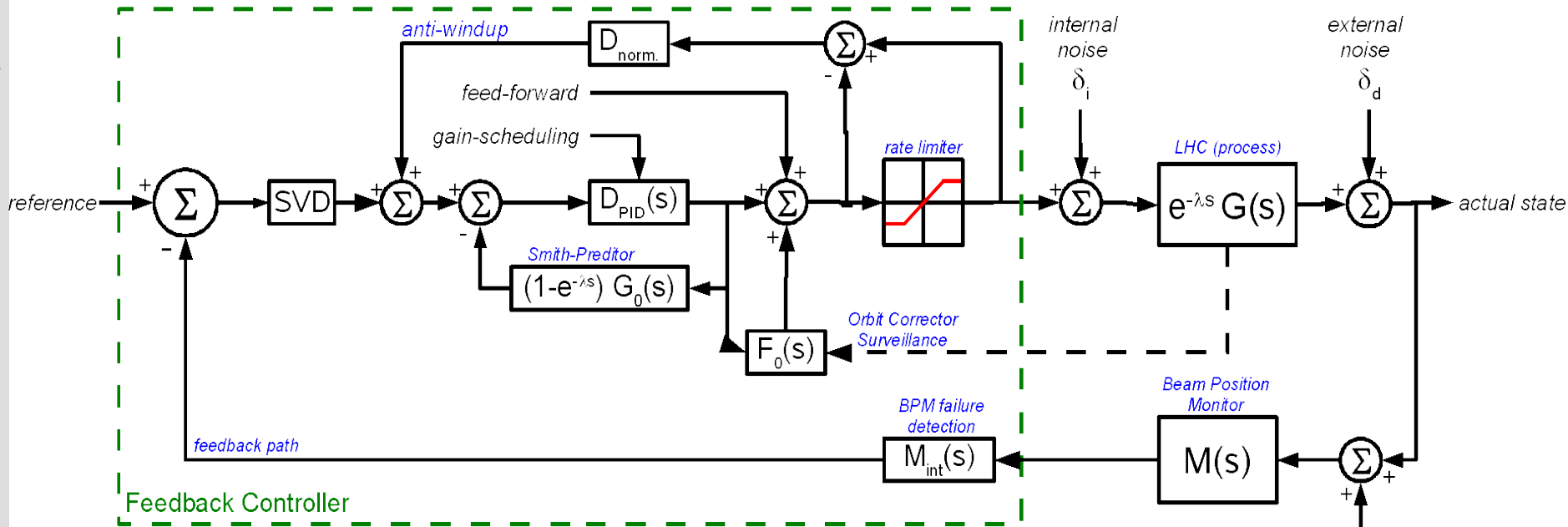


Time Domain: Optimal Controller Design Including Non-Linearities

- If $G(s)$ contains non-stable zeros e.g. delay λ & non-linearities $G_{NL}(s)$

$$G(s) = \frac{e^{-\lambda s}}{\tau s + 1} \cdot G_{NL}(s)$$

- with τ the power converter time constant, then: $G^i(s) = \frac{\tau s + 1}{1}$
- Using (1) and (3) yields $T_0(s) = F_Q(s) \cdot e^{-\lambda s} G_{NL}(s)$
- Inserting in (1) yields Smith-Predictor and Anti-Windup schemes:

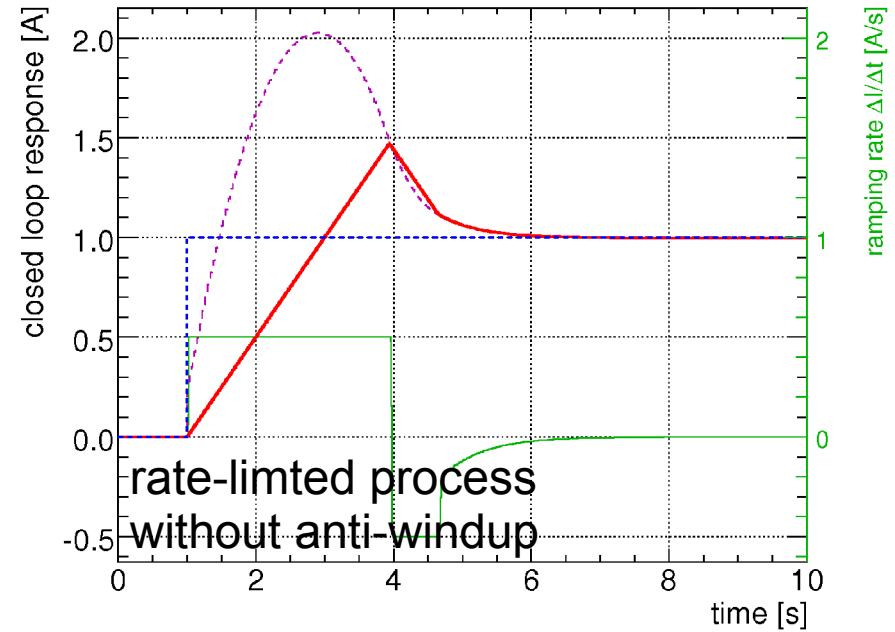
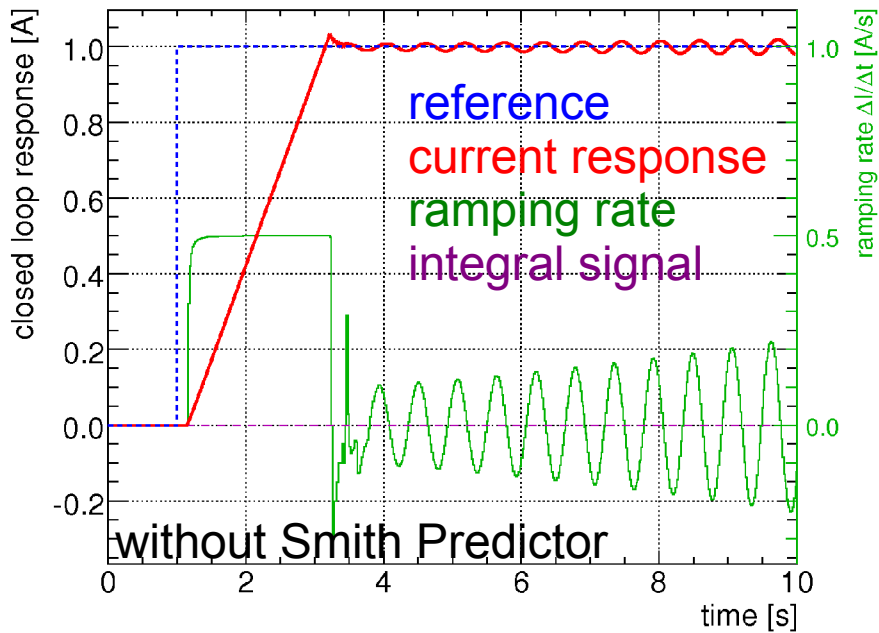


$D_{PID}(s)$ gains are independent on non-linearities and delays!!

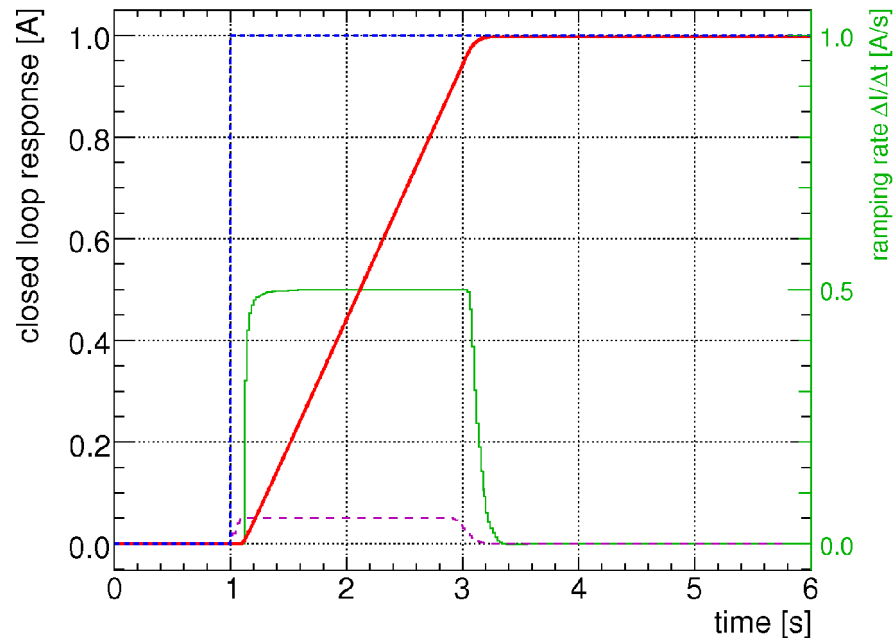
Motivation for Smith-Predictor and Anti-Windup

Example: LHC orbit feedback control

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with full delay and windup compensator scheme:

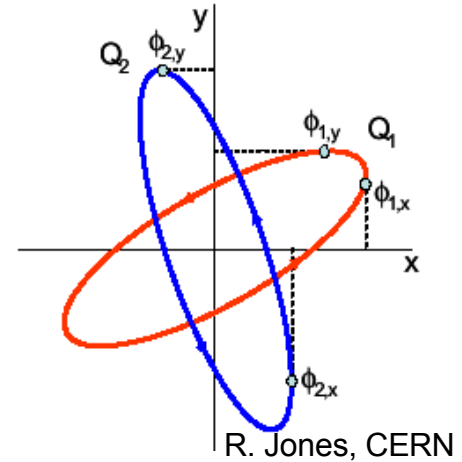
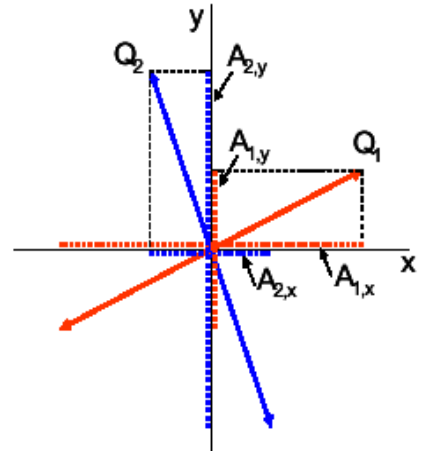


Cross-Dependability and Constrains of FB Loops I/III

Beam Stability Constraints

- Limitation of Orbit to micrometer level imposes constraints on other FBs:
 - Improves machine performance
 - Significantly minimises feed-down effects of higher multipoles (esp. coupling)
 - No (large) kicks/momentum modulation:
 - Q/Q' measurements have to measure within the stability requirements
 - Development of new tune measurement techniques²³
 - CERN SPS/LHC tune PLL operates within $< 1 \mu\text{m}$ excitation level^{21,22}

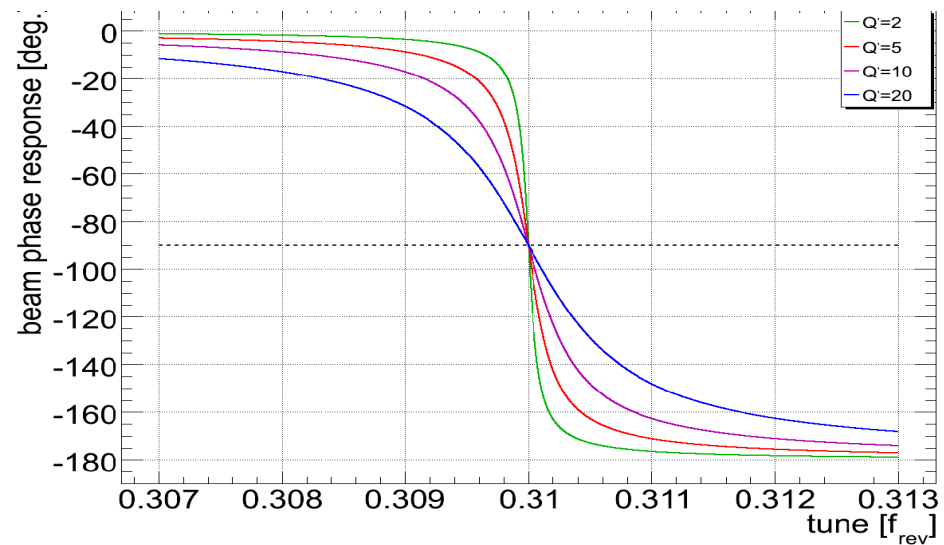
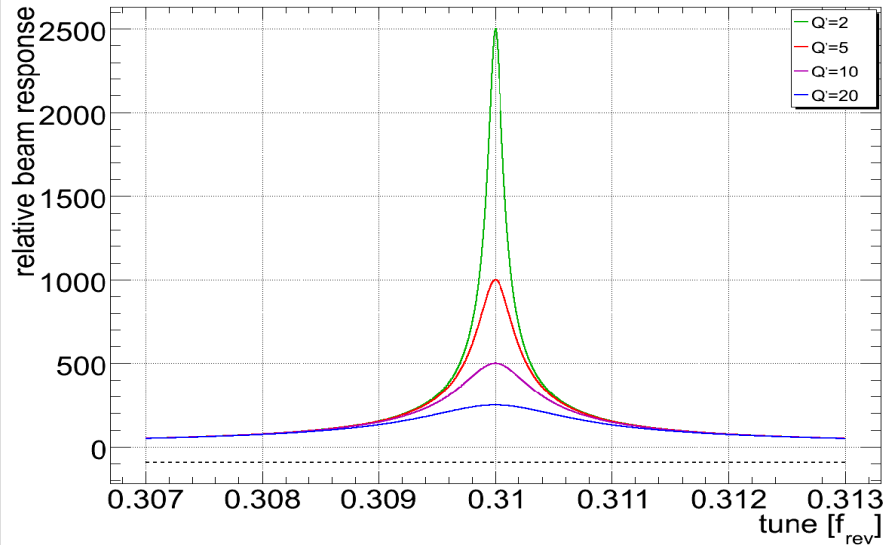
- Robust tune feedback requires measurement/control of Coupling²⁴
 - Additional decoupling between PLL phase and amplitude²²



R. Jones, CERN

Cross-Dependability and Constrains of FB Loops III/III

Tune PLL Gain Dependence on Chromaticity



- Beam response: open loop gain $K_0 \sim$ phase response slope
- Common^{4-6,19} (classic) PLL loop design: $K_0 = \text{const.}$ & filter bandwidth = $1/\tau$
 → PLL low-pass:

$$G(s) = \frac{K_0}{\tau s + 1}$$

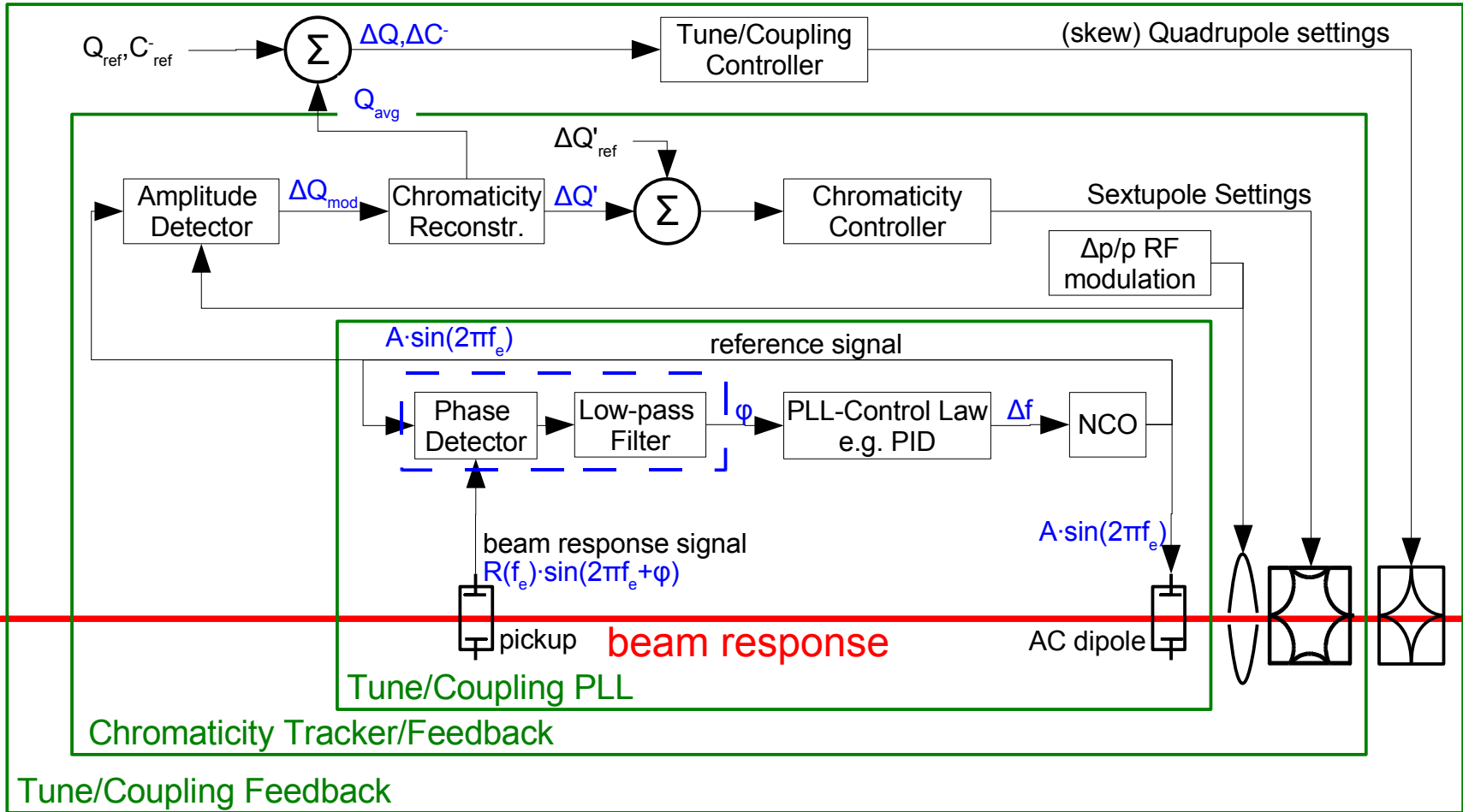
Note:
 K_0 const. for $|\Delta\phi| \leq 60^\circ$ (linear. regime)
 K_0 depends on Q' (non-linear. regime)

- Optimal Tune PLL gain parameter depend on Chromaticity^{20,21}
 - Optimal PI for high Q' ↔ sensitivity to noise (unstable loop) for low Q'
 - Optimal PI for small Q' ↔ slow tracking speed for large Q'

Cross-Dependability and Constrains of FB Loops II/III

Example: Q' FB – 3++ nested and coupled loops

- Traditionally loop designs are often addressed one-by-one
 - neglects cross-dependence and cross-constraints w.r.t. other nested FB loops



- Additional cross-dependence with orbit and energy feedback (dispersion orbit)

- Beam-based FBs are remedies for perturbations on slow/medium time scales
 - Limited by thermal drifts, noise and systematics of involved devices
 - **Systematic and thorough analysis of involved beam instrumentation and corrector circuits is essential!**
- Use of imperfect (design) beam response for SVD based FB Systems:
 - does not affect the precision of the correction but reduces rather the effective bandwidth → favours higher feedback sampling frequencies
- Youla's affine parameterisation facilitates optimal adaptive non-linear control
 - enables gain-scheduling based on operational scenario
 - (Ziegler-Nichols/Coohen-Coon PID tuning are outdated!)
- Beware of cross-constraints/coupling of simultaneous nested loops:
 - Feedbacks should be designed as an ensemble

The author gratefully acknowledges contributions from many colleagues!



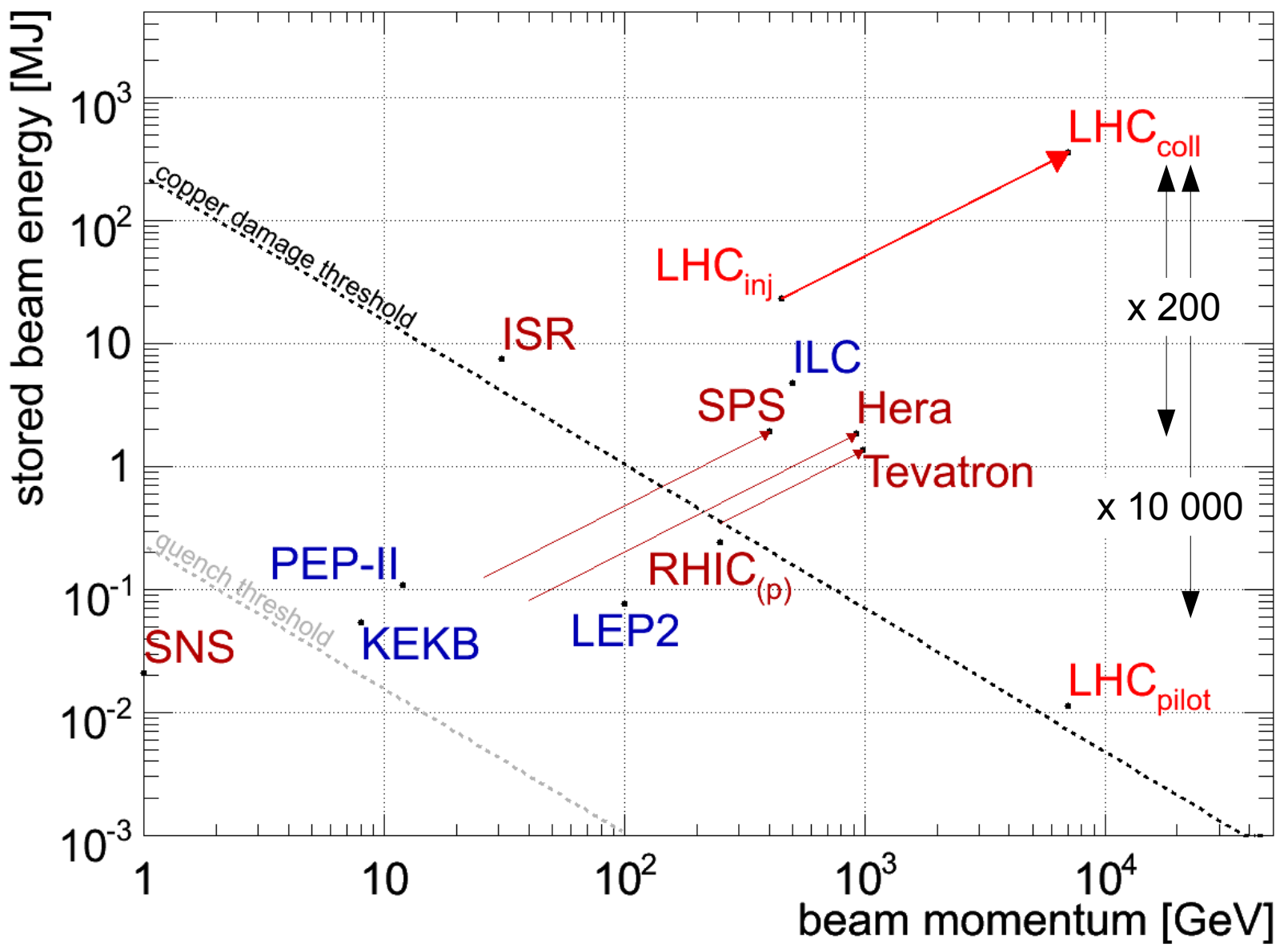
Reserve Slides

- (1) Proceeding of “3rd International Workshop on Beam Orbit stabilization”, 2004, <http://iwbs2004.web.psi.ch/>
- (2) M. Böge, “Achieving Sub-micron Stability in Light Sources”, EPAC'04, Lucerne, 2004
- (3) G Decker, “Beam Stability in Synchrotron Light Sources”, DIPAC'05, Lyon, France, 2005
- (4) O. Berrig et al., “The Q-Loop: A Function Driven Feedback System for the Betatron Tunes During the LEP Energy Ramp”, CERN SL-98-039, 1998
- (5) A.Fisher et al., “Tune Feedback in PEP-II”, SLAC-PUB-10230, 2003
- (6) P. Cameron et al., “Tune Feedback at RHIC”, PAC'01, Chicago, IL, 2001
- (7) P. Cameron et al., “Measuring Chromaticity Along the Ramp Using the PLL Tune-Meter in RHIC”, EPAC'02, Paris, 2002
- (8) R. Hettel, “Beam Stability Issues at Light Sources”, 25th ICFA Advanced Beam Dynamics Workshop, Shanghai, 2002
- (9) Chamonix XIV, “Damage levels - Comparison of Experiment and simulation” and PAC'05
- (10) R. Assmann, “Collimation and Cleaning: Could this limit the LHC Performance?”, Chamonix XII, 2003
- (11) S. Redaelli, “LHC aperture and commissioning of the Collimation System”, Chamonix XIV, 2005
- (13) R. Steinhagen, “Closed Orbit and Protection”, CERN, MPWG #53, 2005-12-16
- (14) S. Fartoukh, O. Brüning, “Field Quality Specification for the LHC Main Dipole Magnets”, LHC Project Report 501
- (15) S. Fartoukh, J.P. Koutchouk, “On the Measurement of the Tunes, [...] in LHC”, LHC-B-ES-0009, EDMS# 463763
- (16) G. Golub and C. Reinsch, “*Handbook for automatic computation II, Linear Algebra*”, Springer, NY, 1971
- (17) D. C. Youla et al., “*Modern Wiener-Hopf Design of Optimal Controllers*”, IEEE Trans. on Automatic Control, 1976, Vol. 21-1, pp. 3-13 & 319-338
- (18) B. Anderson, “From Youla-Kucera to Identification, Adaptive and Non-linear Control”, Automatica, 1998, Vol. 34, No. 12, pp. 1485-1506
- (19) C.Y. Tan, “The Tevatron Tune Tracker PLL”, FERMILAB-TM-2275-AD

- (20) R.J. Steinhagen et al., “Influence of Varying Chromaticity on Robustness of the LHC Tune PLL and its Application for Continuous Chromaticity Measurement.”, PAC'07, 2007
- (21) R.J. Steinhagen, “First Results of the PLL Tune Tracking in the SPS”, Tune Feedback Final Design Review, BNL, http://www.agsrhichome.bnl.gov/LARP/061024_TF_FDR/index.html
- (22) R.J. Steinhagen et al., “The LHC Phase-Locked-Loop for Continuous Tune Measurement - Prototype tests at the CERN-SPS”, PAC, 2007
- (23) M. Gasior, R. Jones, “The principle and first results of betatron tune measurement by direct diode detection”, CERN-LHC-Project-Report-853
- (24) R. Jones et al., “Towards a Robust Phase Locked Loop Tune Feedback System”, DIPAC'05, Lyon, France, 2005
- (25) T. Schilcher, “Commissioning and Operation of the SLS Fast Orbit Feedback”, EPAC'04, Lucerne/Switzerland, 2004
- (26) K. Park, “Development of High Precision Magnet Power Supply using the DSP”, this proceedings
- (27) Q. King, “The All-Digital Approach to LHC Power Converter Current Control”, ICALEPS'01, 2001
- (23) R.J. Rushton, “Diamond Storage Ring Power Converters”, EPAC'06, Edinburgh, Scotland, 2006
- (29) P. Gros, “The 3Hz Power Supplies of the SOLEIL Booster”, EPAC'06, Edinburgh, Scotland, 2006

IWBS'04: "LHC is a pretty dangerous machine" Livingston Style plot

APAC'07, Real-Time Feedback on Beam Parameters, Ralph.Steinhagen@CERN.ch, 2007-01-29

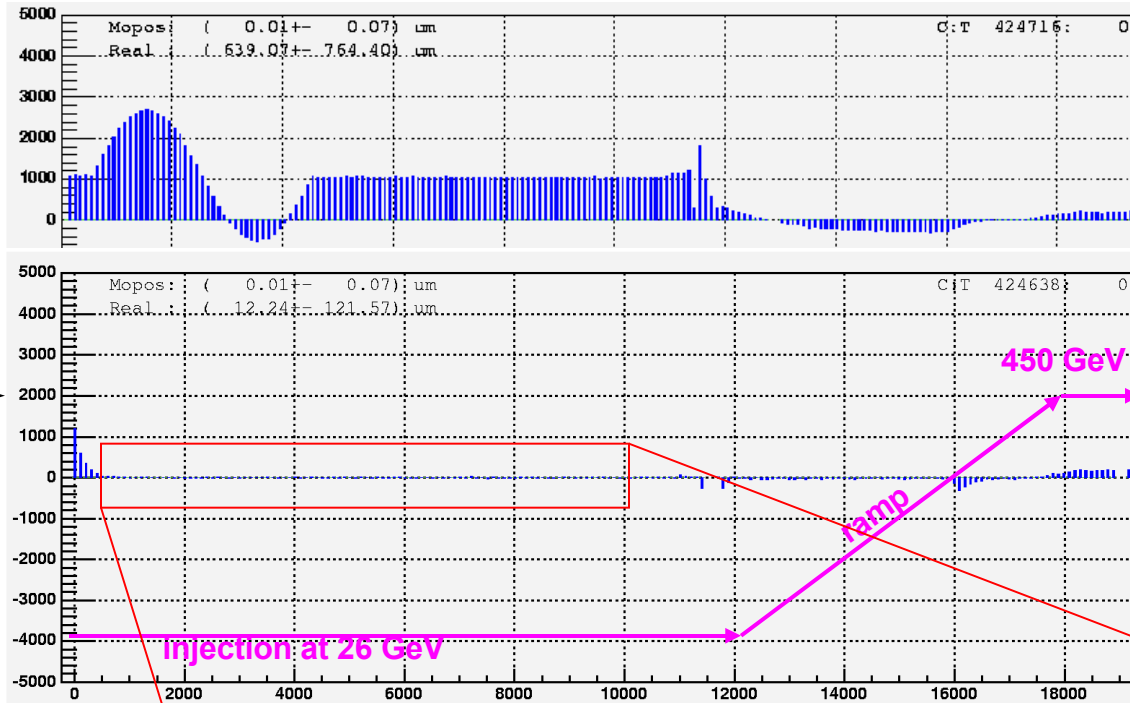


see Chamonix XIV: "Damage levels - Comparison of Experiment and simulation" and PAC'05 for details

LHC Orbit Feedback Test at the SPS I/II

BPM
Reading
(μm)

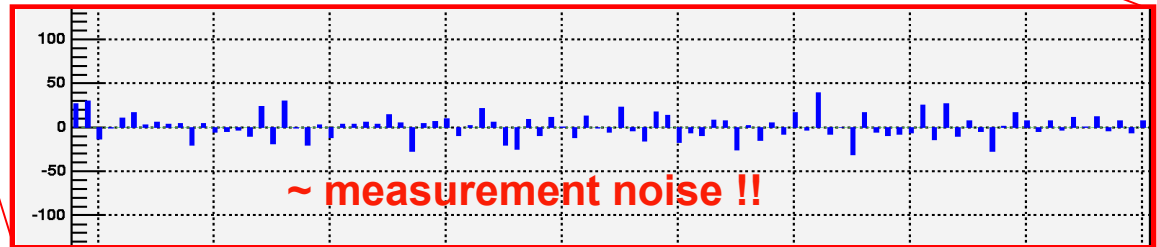
Time (ms)



feedback off

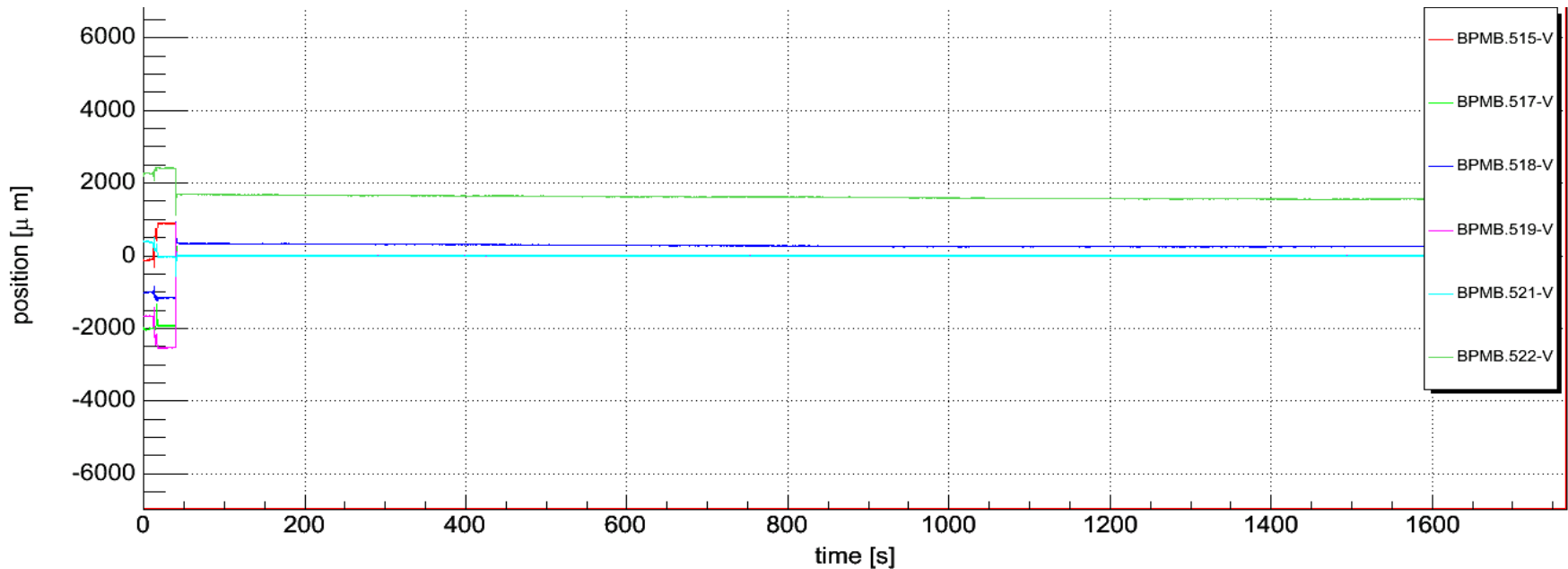
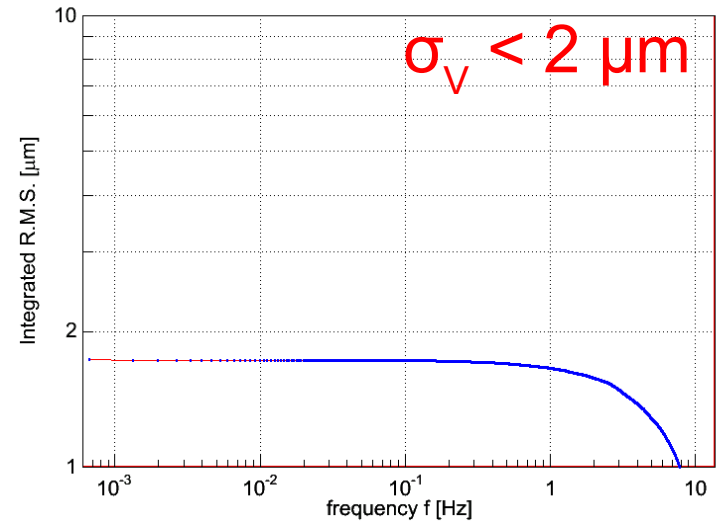
feedback on

feedback on (zoom)



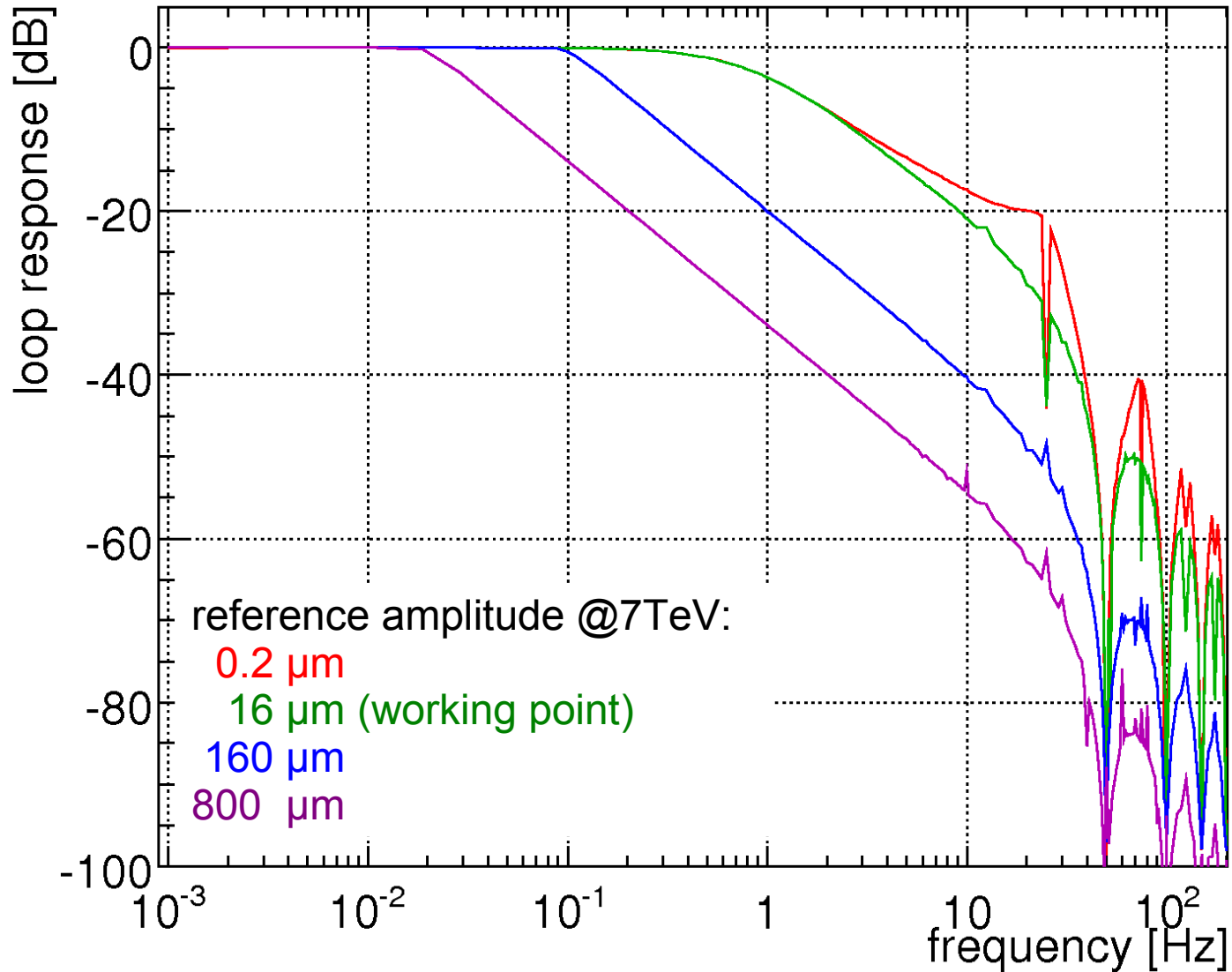
LHC Orbit Feedback Test at the SPS III/II

- Stabilisation “record” in the SPS
 - 270 GeV coasting (proton) beam, 72 nom bunches, $\beta_v \approx 100$ m
 - rivals most modern light sources
 - magnitudes better than required
- Target: maintain same longterm stability



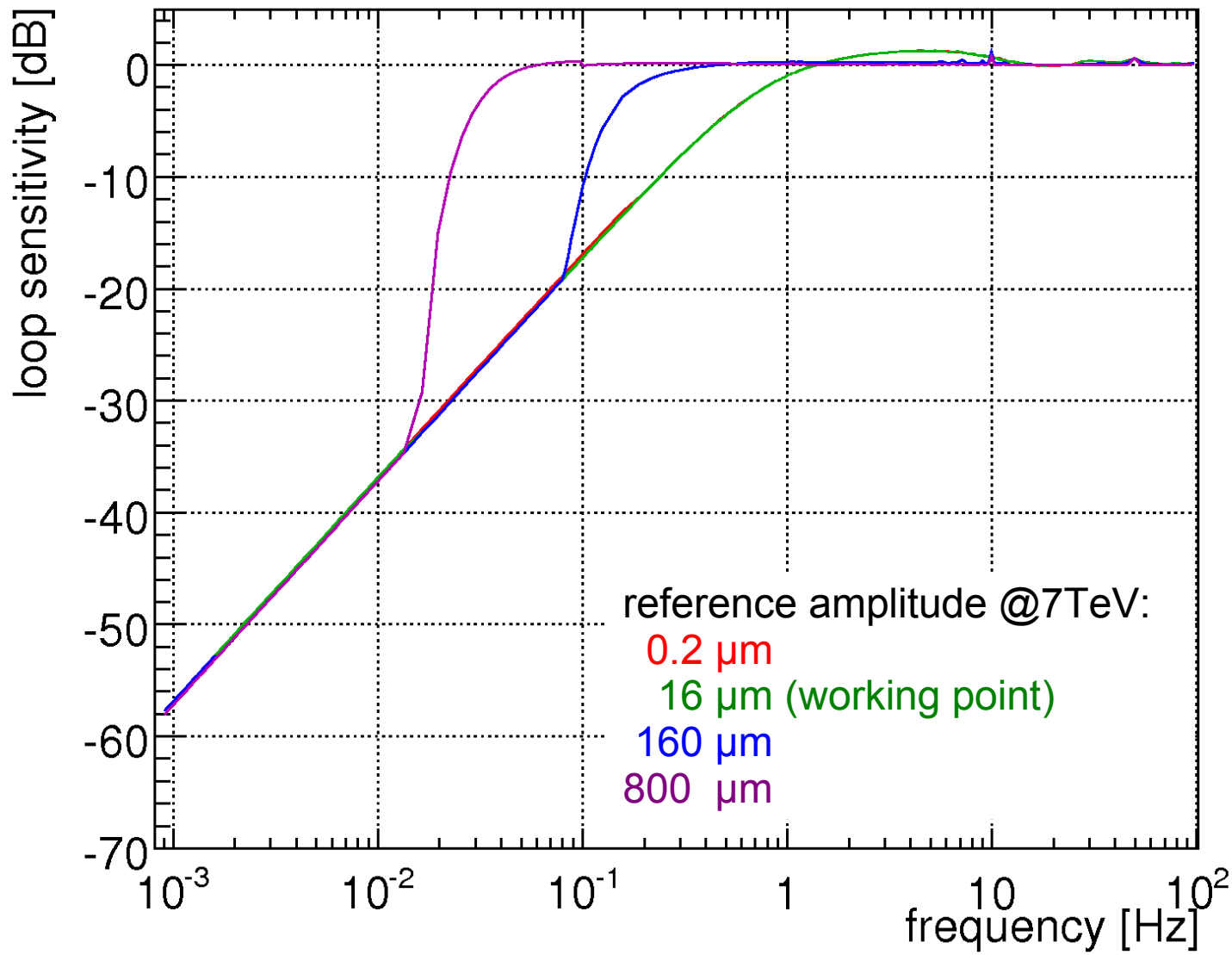
Nominal Feedback Response T_0

- Full LHC orbit simulation @1KHz sampling, (BPM sampling: 25Hz)



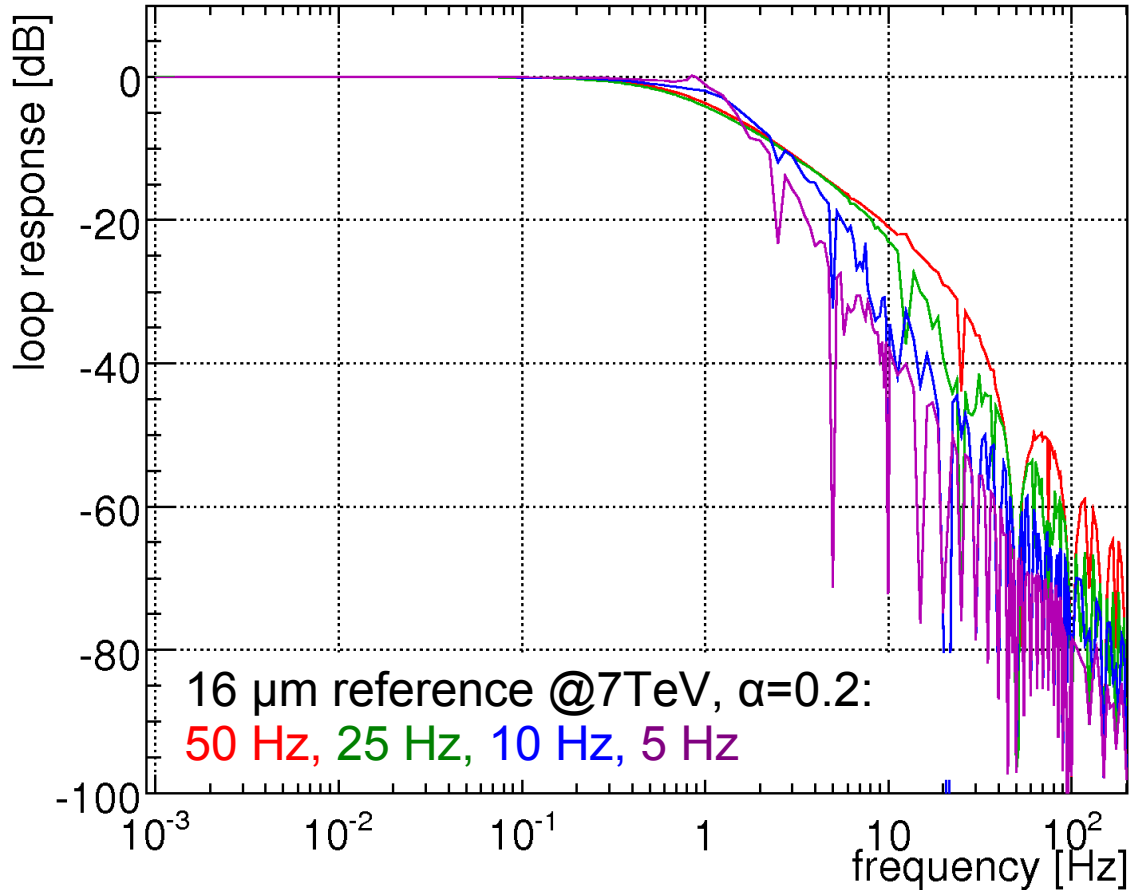
Nominal Feedback Disturbance Rejection S_{d0}

- Full LHC orbit simulation @1KHz sampling, (BPM sampling: 25Hz)



Loop Bandwidth versus Sampling frequency

- ... sample the position (Q, ...) at 10Hz to achieve a closed loop 1Hz bandwidth

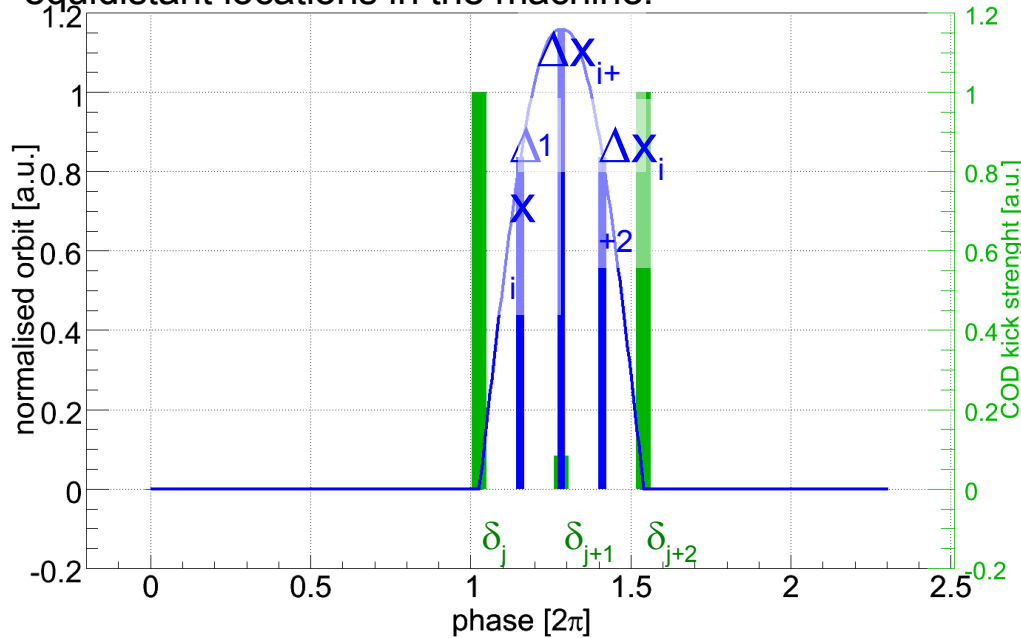


- ... a theoretic limit assuming a perfect system!
- common: sampling frequency $> 25 \dots 40$ desired closed-loop bandwidth

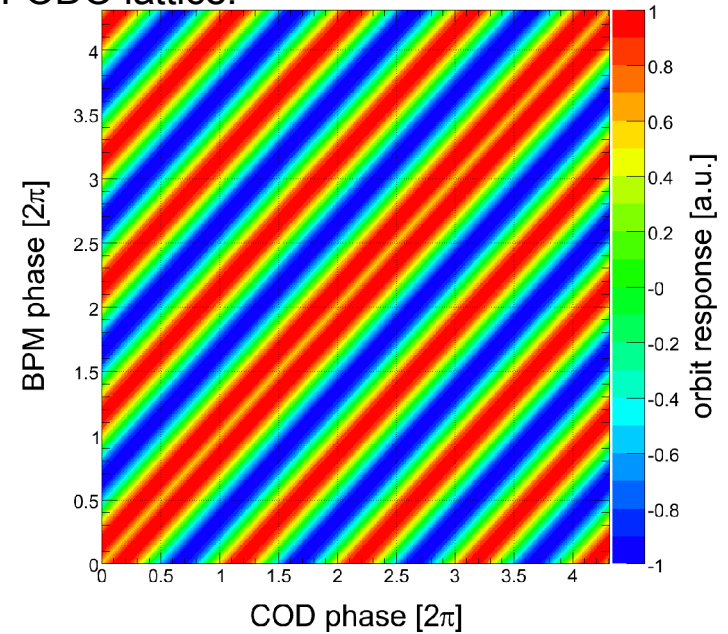


Automated Orbit Correction using Singular Value Decomposition

The orbit is sampled at m discrete not necessarily equidistant locations in the machine:



orbit response matrix example of a regular FODO lattice:



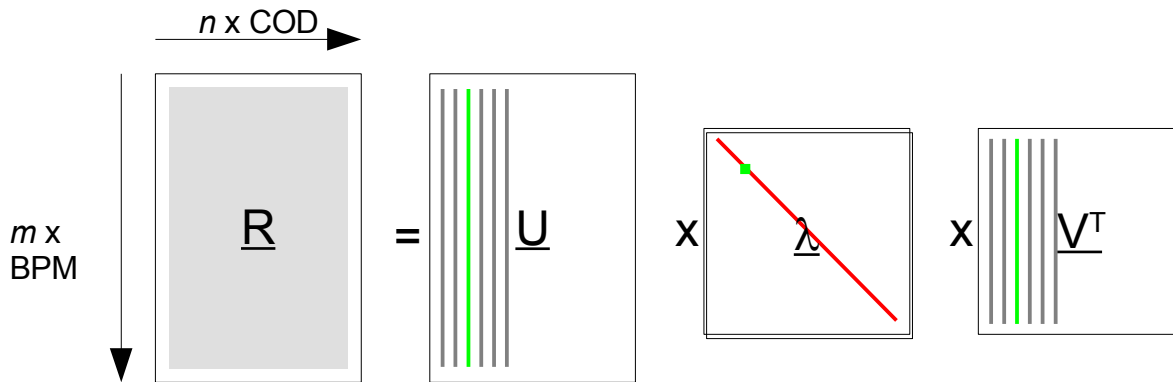
The superimposed beam position shift at the i^{th} monitor due to single dipole kicks is described through the orbit response matrix \underline{R} . It can be written as

$$\Delta x_i = \sum_{j=0}^n R_{ij} \cdot \delta_j \quad \text{with} \quad R_{ij} = \frac{\sqrt{\beta_i \beta_j}}{2 \sin(\pi Q)} \cdot \cos(\Delta \mu_{ij} - \pi Q)$$

$$\Leftrightarrow \Delta \vec{x} = \sum_{j=0}^n \delta_j \vec{u}_j \quad \text{with} \quad \vec{u}_j = (R_{1j}, \dots, R_{mj})^T \Leftrightarrow \Delta \vec{x}(t) = \underline{R} \cdot \vec{\delta}_{ss}$$

where (β, μ, Q) depends on the machine optic (example: $Q=4.31$).

Theorem from linear algebra*:



eigen-vector relation:

$$\lambda_i \vec{u}_i = \underline{R} \cdot \vec{v}_i$$

$$\lambda_i \vec{v}_i = \underline{R}^T \cdot \vec{u}_i$$

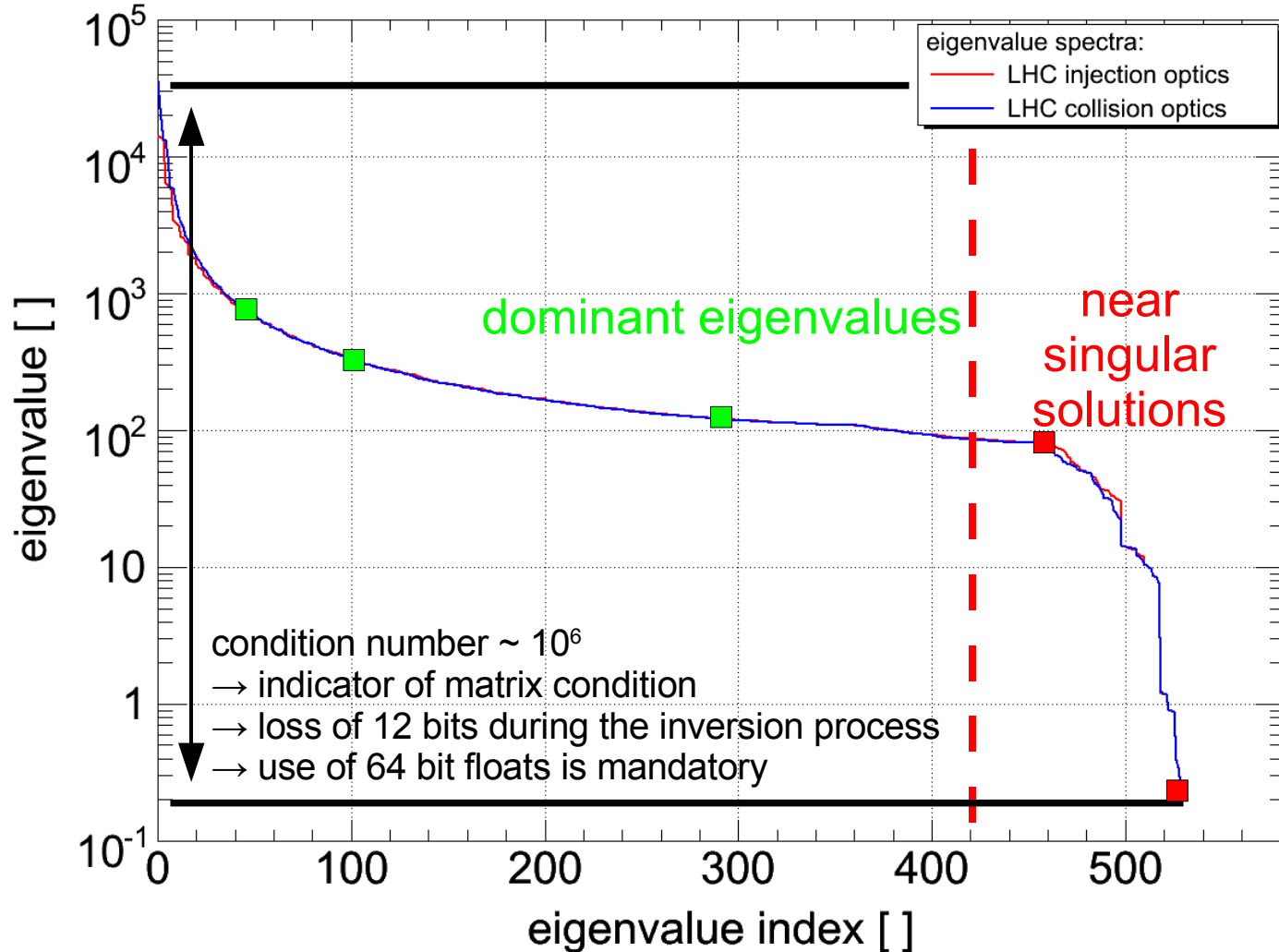
- final correction is a simple matrix multiplication
- large eigenvalues \leftrightarrow bumps with small COD strengths but large effect on orbit

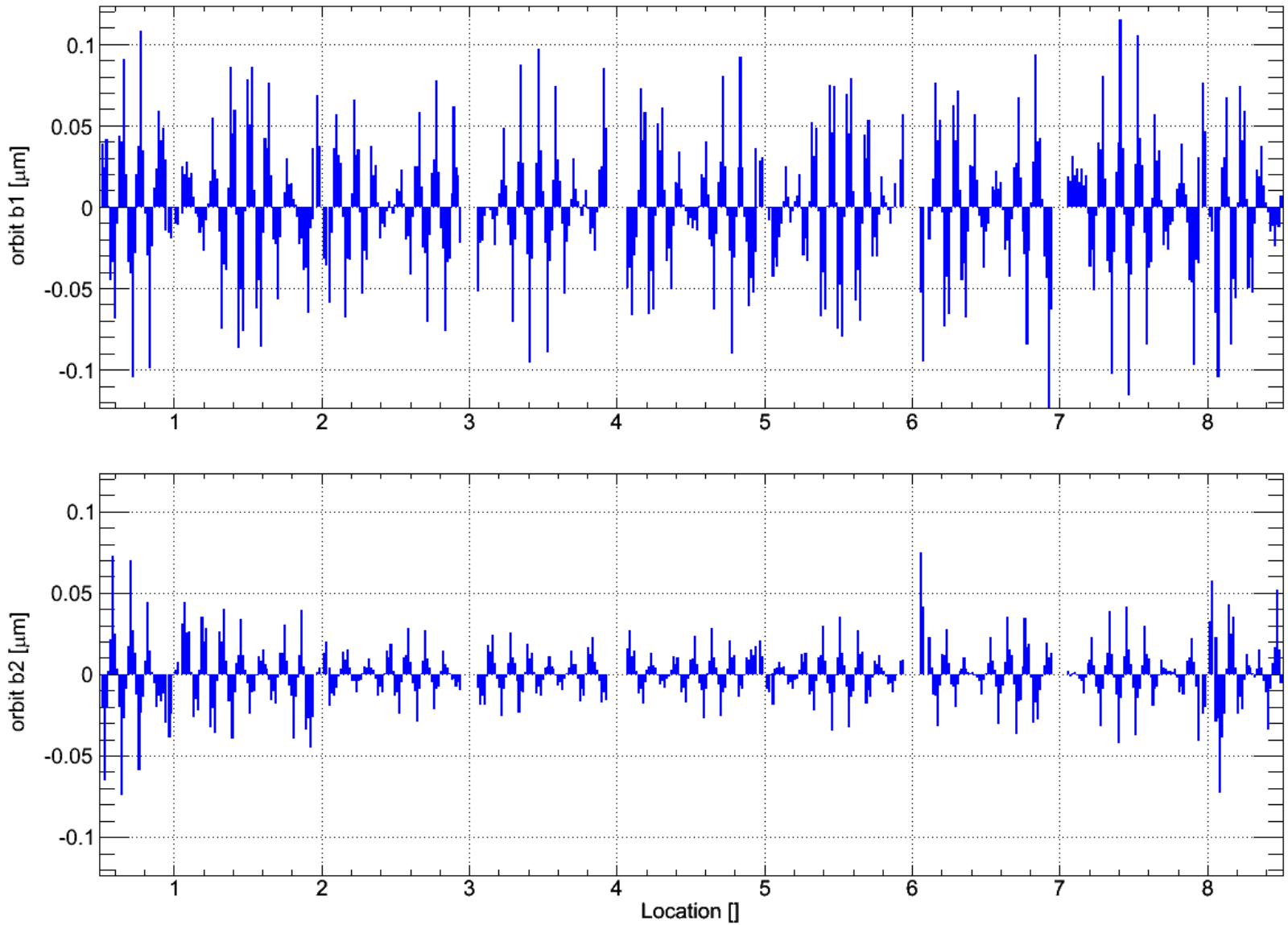
$$\vec{\delta}_{ss} = \tilde{R}^{-1} \cdot \Delta \vec{x} \quad \text{with} \quad \tilde{R}^{-1} = \underline{V} \cdot \underline{\Lambda}^{-1} \cdot \underline{U}^T \quad \Leftrightarrow \quad \vec{\delta}_{ss} = \sum_{i=0}^n \frac{a_i}{\lambda_i} \vec{v}_i \quad \text{with} \quad a_i = \vec{u}_i^T \Delta \vec{x}$$

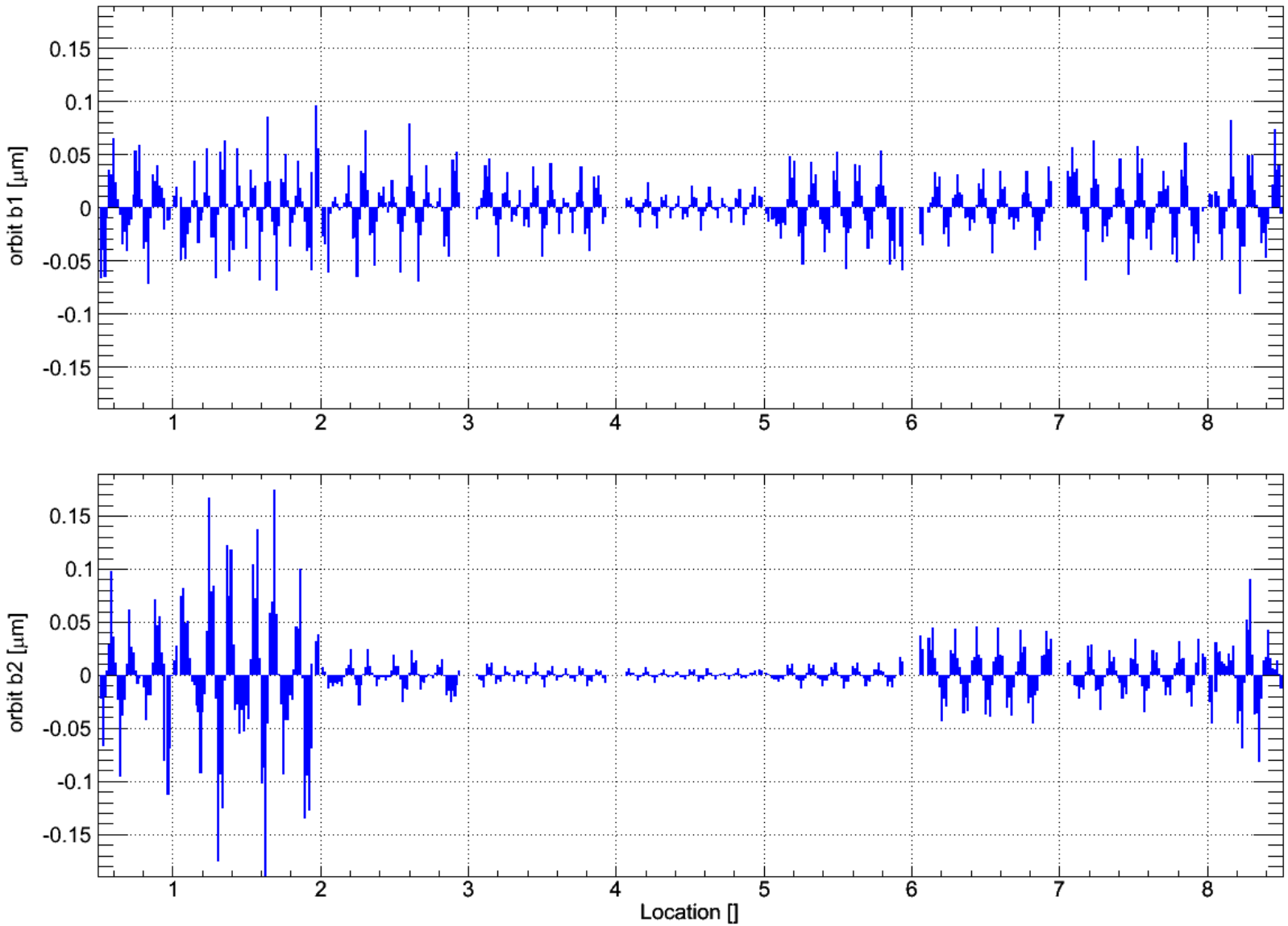
- Easy removal of singular (=undesired, large corrector strengths) eigen-values/solutions:
 - near singular eigen-solutions have $\lambda_i \sim 0$ or $\lambda_i = 0$
 - to remove those solution: $\lim_{\lambda_i \rightarrow \infty} 1/\lambda_i = 0$
- **discarded eigenvalues corresponds to bumps that won't be corrected by the fb**

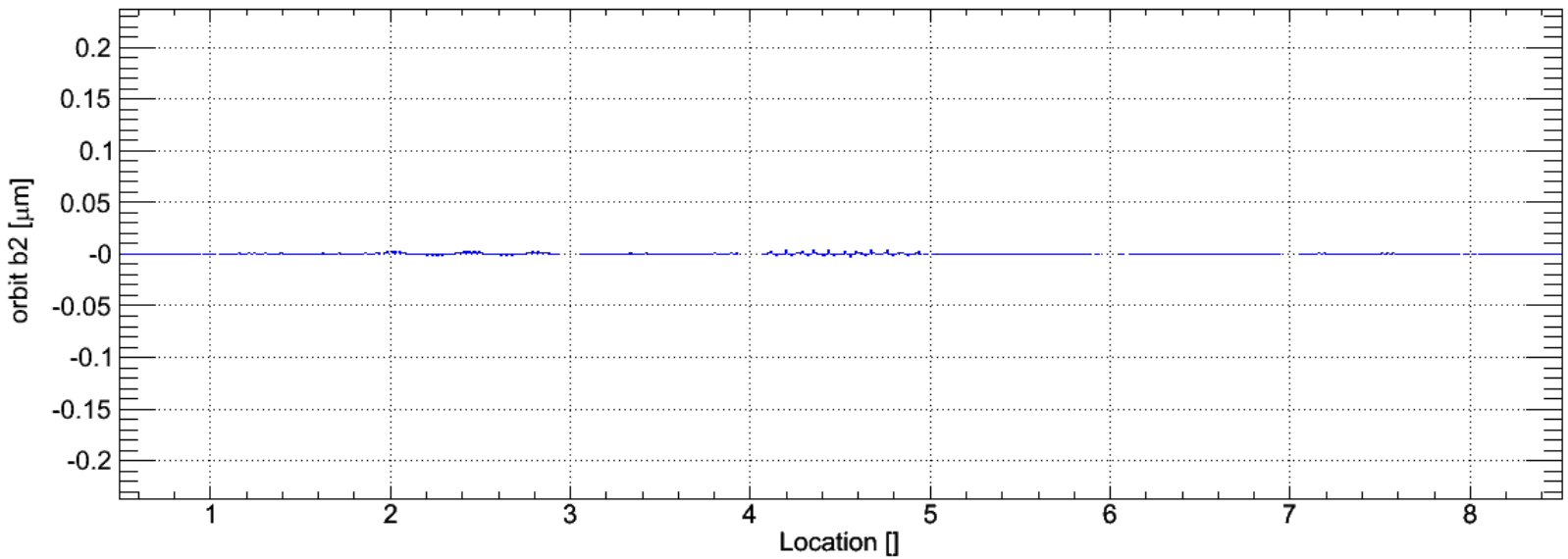
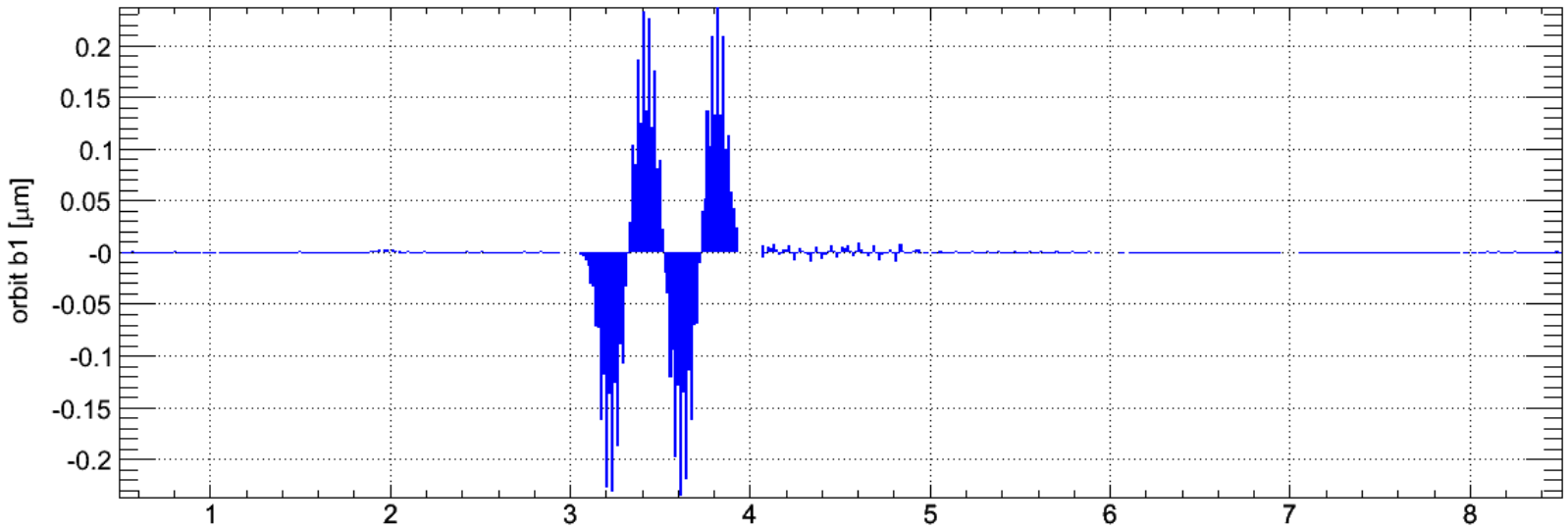
*G. Golub and C. Reinsch, "Handbook for automatic computation II, Linear Algebra", Springer, NY, 1971

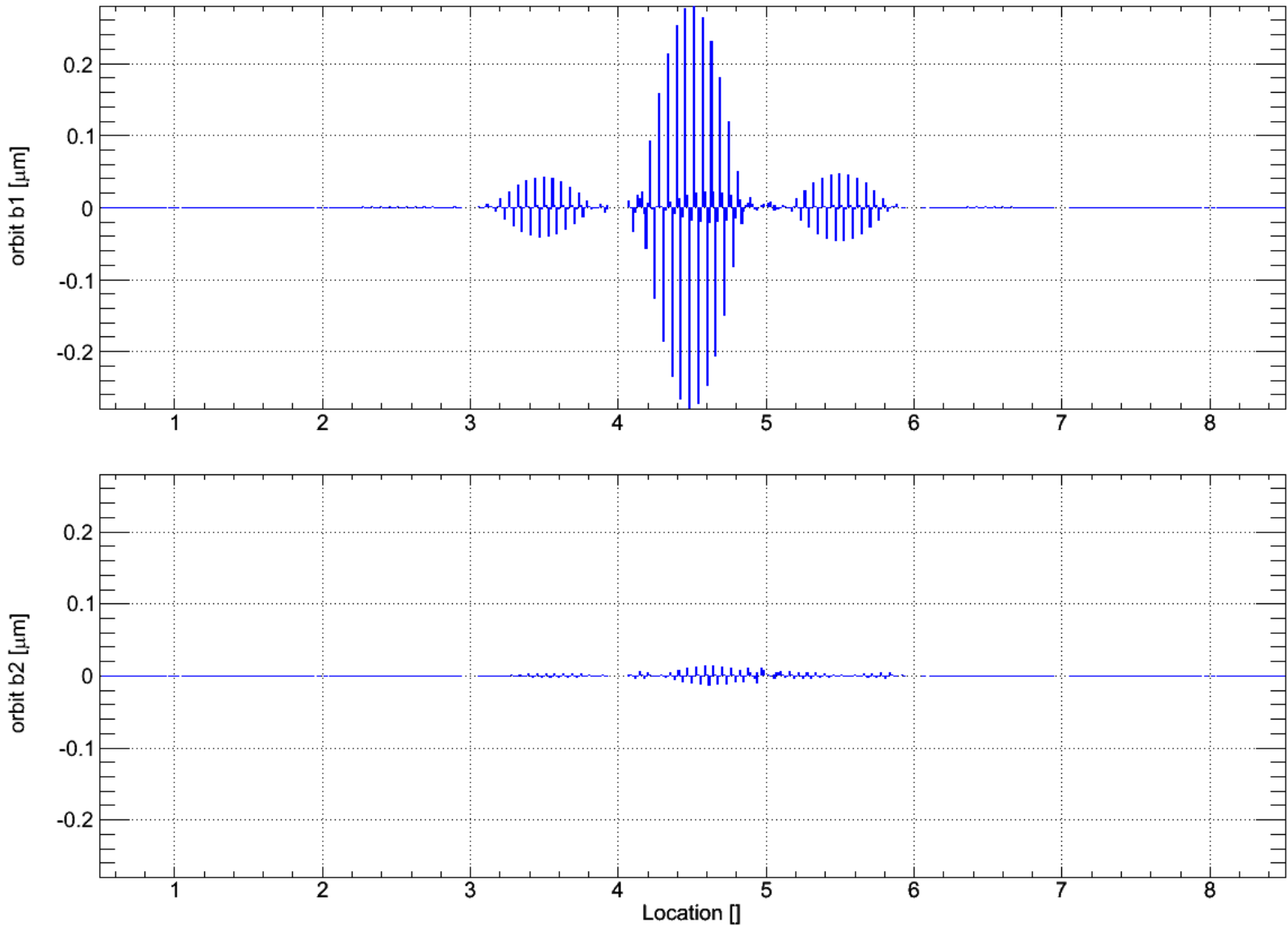
Eigenvalue spectra for vertical LHC response matrix using all BPMs and CODs:

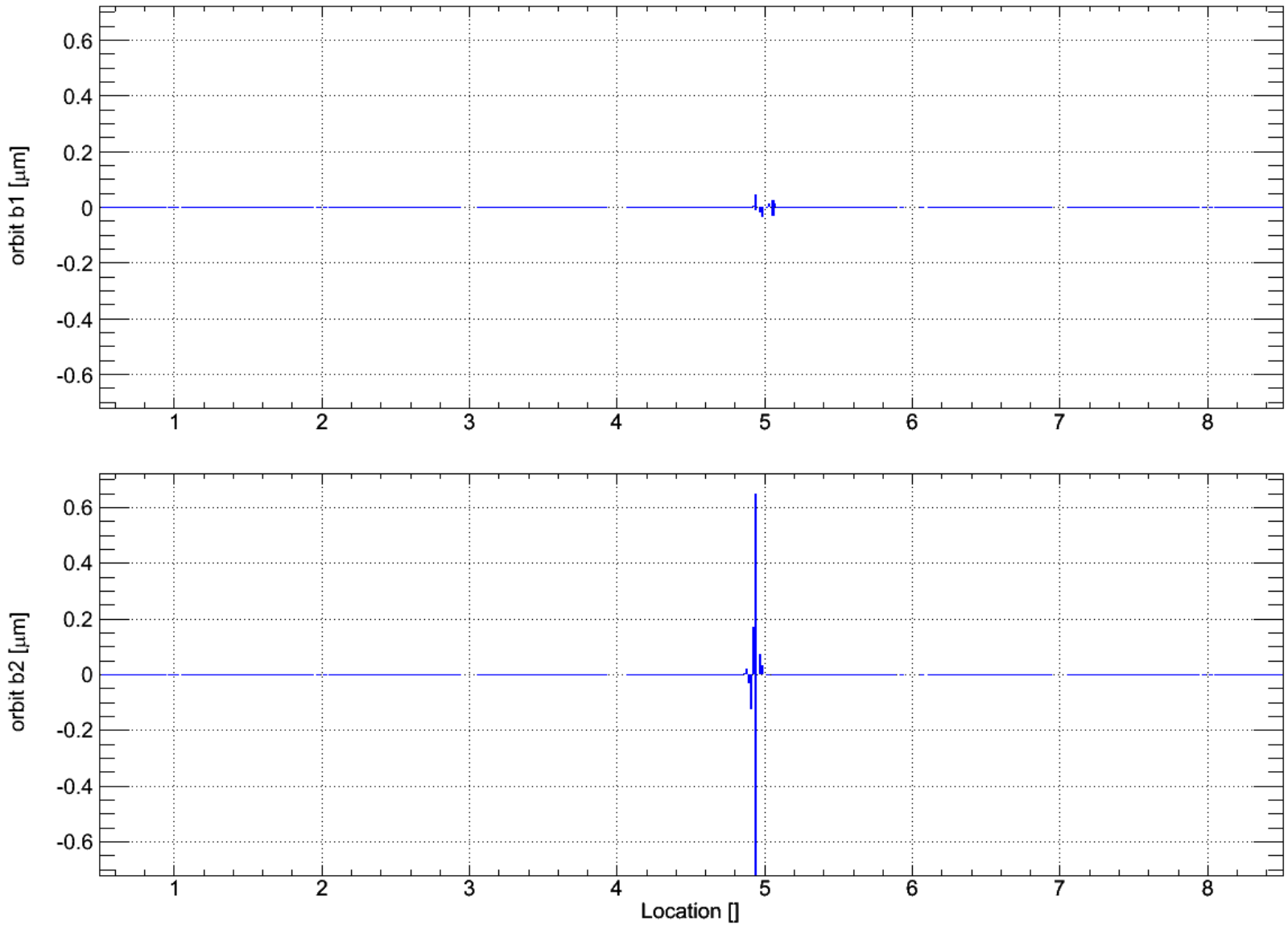




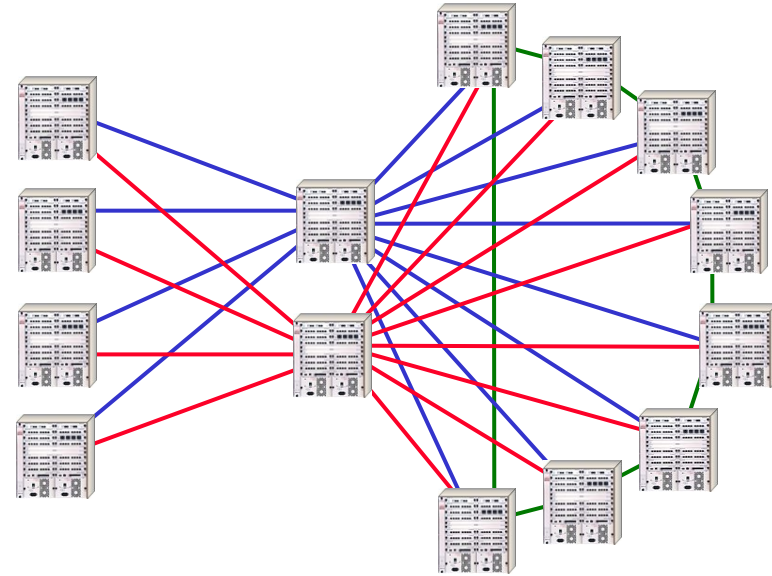






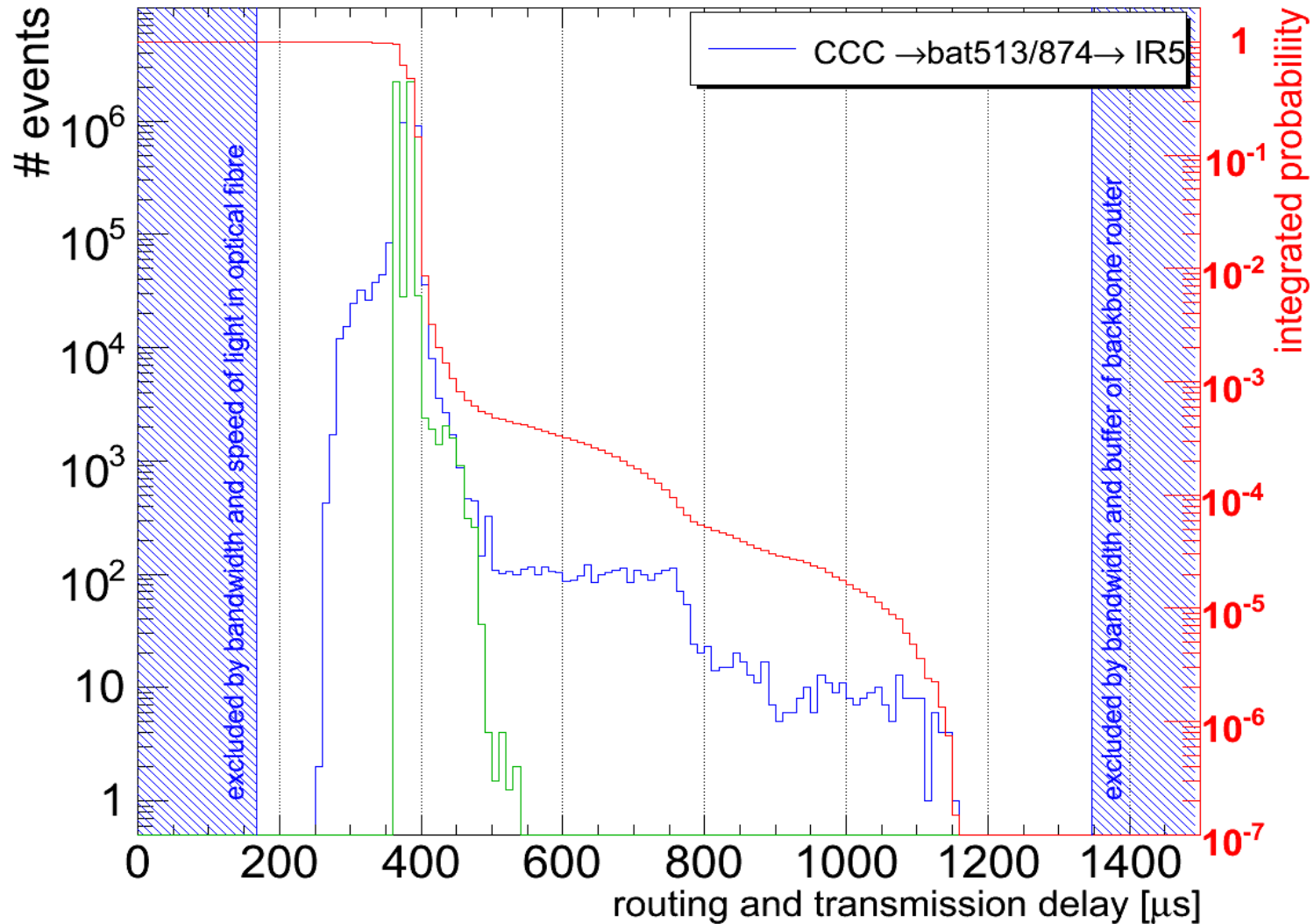


- CERN's Technical Network as backbone
 - Switched network
 - no data collisions
 - no data loss
 - double (triple) redundancy
- Core: “Enterasys X-Pedition 8600 Routers”
 - 32 Gbits/s non-blocking, $3 \cdot 10^7$ packets/s
 - 400 000 h MTBF
 - hardware QoS
 - One queue dedicated to real-time feedback
 - ~ private network for the orbit feedback



- Routing delay ~ 13 μ s
- longest transmission delay (exp. verified) ~ 320 μ s
(500 bytes, IP5 -> Control room ~5 km)
 - 20% due to infrastructure (router/switches)
 - 80% due to traveling speed of light inside the optic fibre
- worst case max network jitter « targeted feedback frequency!

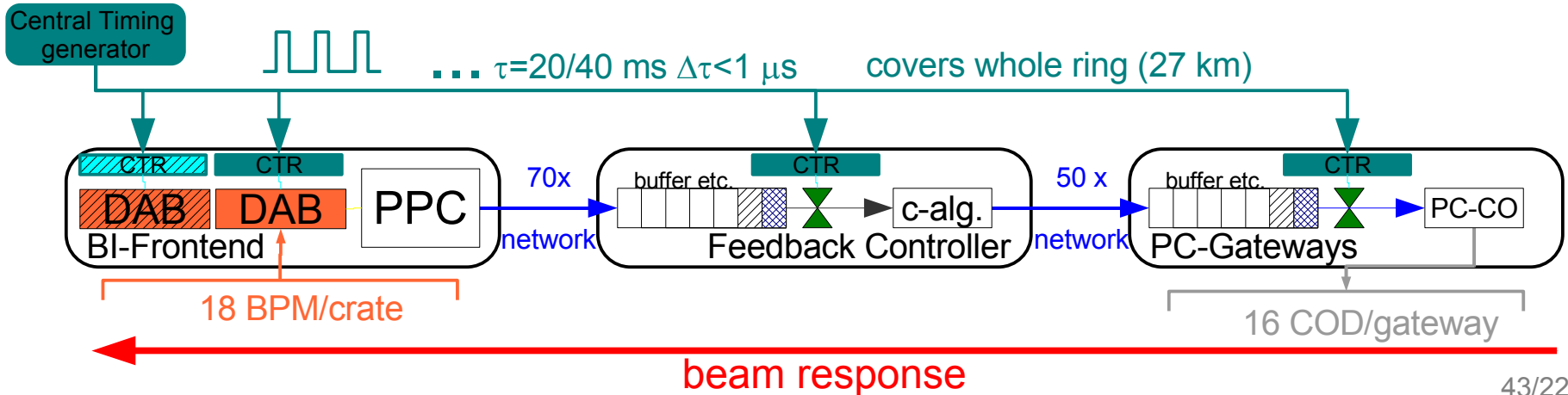
- The maximum latency between CCC and IR5
 - tail of distribution is given by front-end computer and its operating system



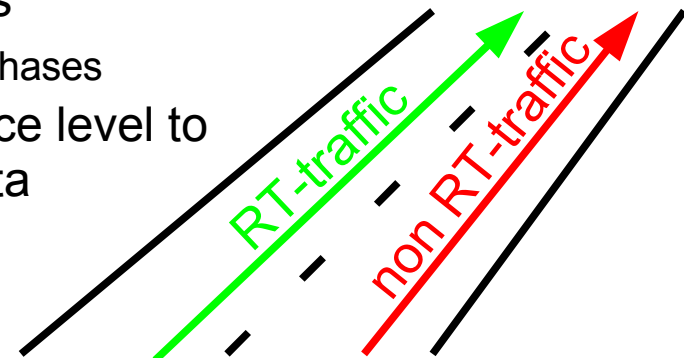
Remaining Jitter Compensation: Fix Max Loop Delay

Two main strategies:

- actual delay measurement and dynamic compensation in SP-branch:
 - high numerical complexity, due to continuously changing branch transfer function
 - only feasible for small systems
- Jitter compensation using a periodic external signal:
 - CERN wide synchronisation of events on sub ms scale that triggers:
 - Acquisition of BPM system, reading of receive buffers, processing and sending of data, time to apply in the power converter front-ends
 - The total jitter, the sum of all worst case delays, must stay within “budget”.
 - Measured and anticipated delays and their jitter are well below 20 ms.
 - feedback loop frequency of 50 Hz feasible for LHC, if required...



- The front-end **network interfaces** are presently the bottleneck. e.g. feedback controller @ 50 Hz:
- lots of in-/outbound connections:
 - Two types of loads:
 - Real-Time: BPM and COD control data
 - Avg. bandwidth: ~13 Mbit/s
 - short bursts: **full I/O load within few ms**
(100 MBit/s resp. 1GBit/s, burst duration desired to be short in order to minimise the total loop delay)
 - Non-Real-Time:
 - transfer of new settings to OFC (matrix ~30 MB)
 - PID configuration etc.
 - relay of BPM and feedback data (monitoring/logging)
 - ...
 - (Peak) load similar to high-end network servers
 - Nearly constant full load during certain operational phases
- **network interface** should be scheduled on the device level to provide a **Quality of Service** (QoS) for real-time data
 - One **reserved FIFO** queue for feedback data
 - **General purpose** queue for other data

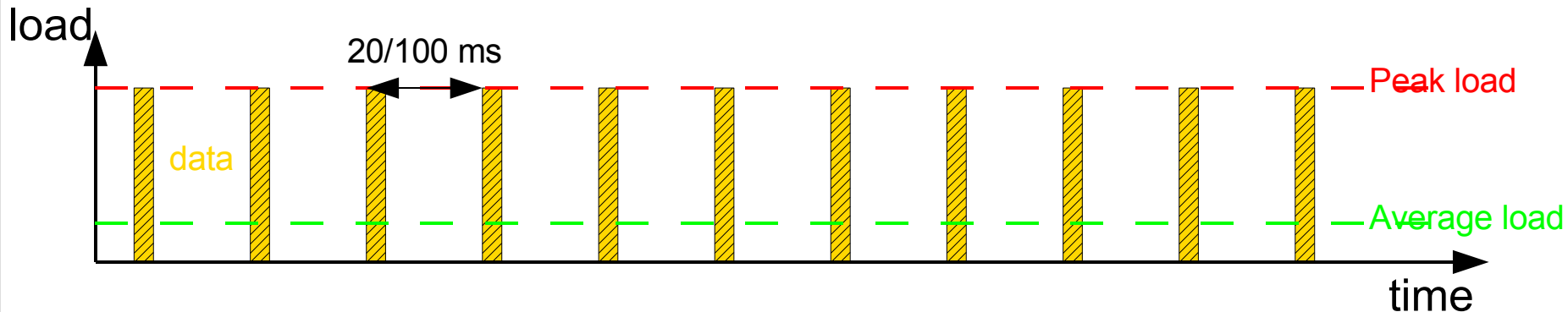


Hardware:

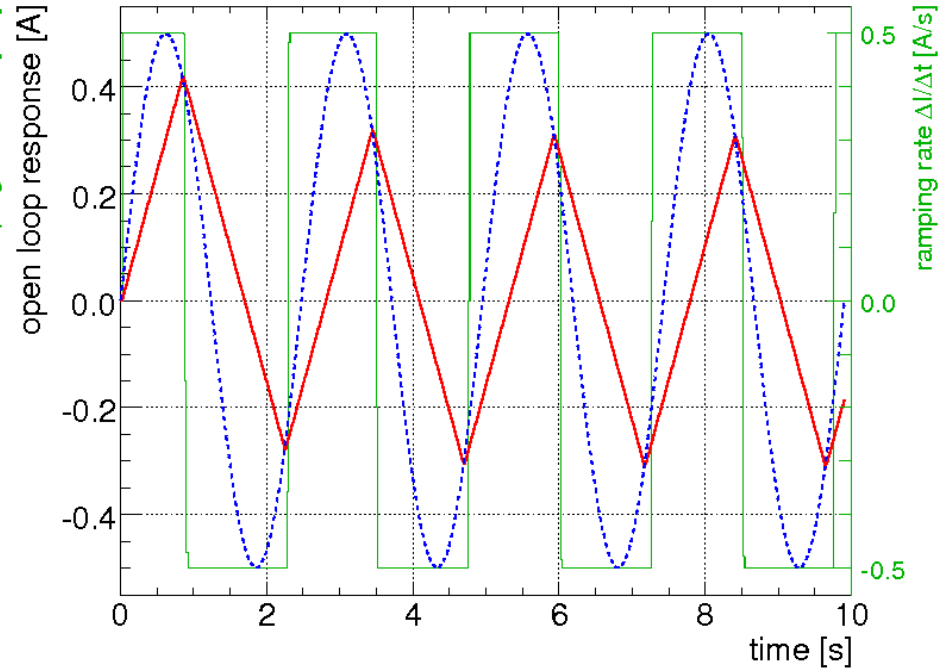
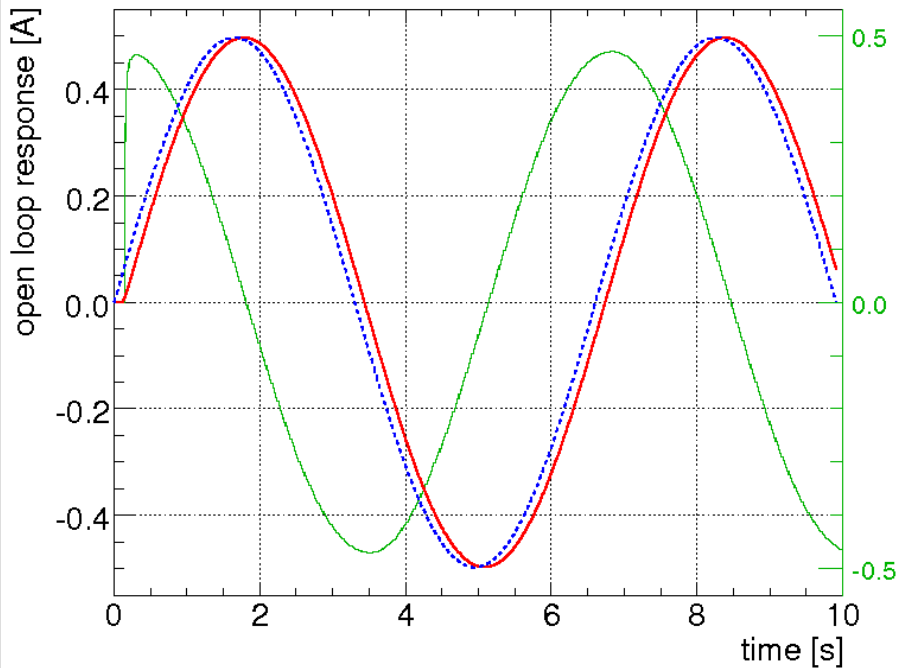
- both rings covered by **1056 BPMs**
- Measure both planes (2112 readings)
- Organised in front-end crates (PowerPC/VME) in surface buildings
 - 18 BPMs (hor & vert) \Leftrightarrow 36 positions / VME crate
 - 68 crates in total, 6-8 crates /IR

Data streams:

- Average** data rates per IR:
 - 18 BPMs x 20 bytes+overhead \sim 1500 bytes / sample / crate
 - 1056 BPMs x 20 byte \sim 94 kbytes / sample
 - @ 10 Hz: \sim 7.7 Mbit/s
 - @ 50 Hz: \sim **38.4 Mbit/s**
- Peak** data rates (bursts): 100Mbit/s resp. 1Gbit/s (depending on Ethernet interface)

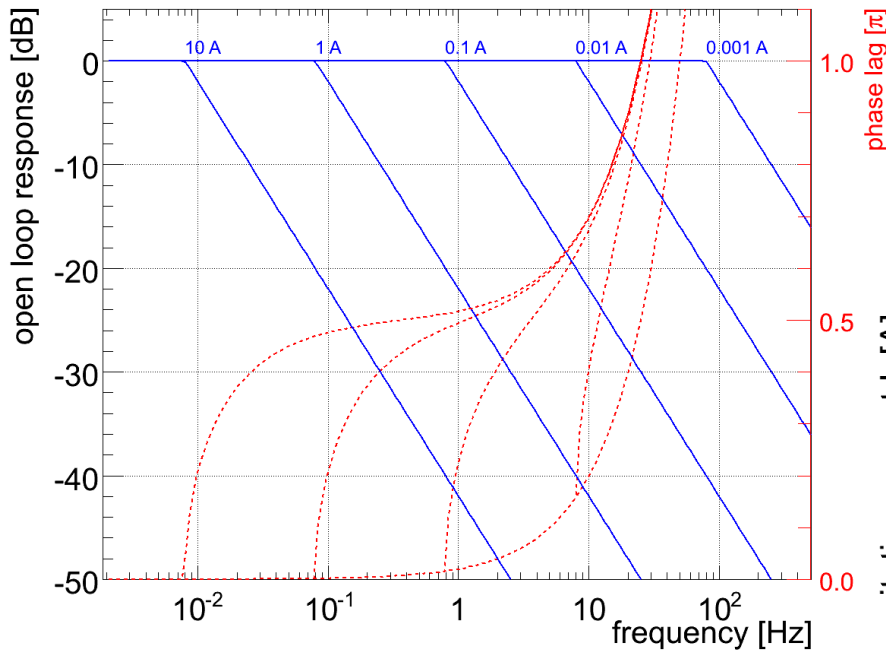


- Two main dynamic contributions
 - Delays: computation, data transmission, etc.
 - Slew rate of the corrector circuits (voltage limitation):
 - $\pm 60\text{A}$ converter: $\Delta I/\Delta t|_{\text{max}} < 0.5 \text{ A/s}$
 - $\pm 600\text{A}$ converter: $\Delta I/\Delta t|_{\text{max}} < 10 \text{ A/s}$



Including Non-Linearities in the Controller Design I/III

- The open-loop corrector circuit bandwidth depends on the excitation current:
 - non-linear phase once rate limiter is in action



$\Delta I = 0.1 \text{ A} \leftrightarrow \Delta x \approx 16 \text{ } \mu\text{m} @ \beta = 180 \text{ m}$

- Consider $\sim 16 \mu\text{m} @ 1 \text{ Hz}$ as effective bandwidth @ 7 TeV
- Injection: ~ 15 times faster!

